



UNIVERSIDAD POLITÉCNICA DE MADRID
Escuela Técnica Superior de Ingenieros Industriales
DIVISIÓN DE INGENIERÍA ELECTRÓNICA

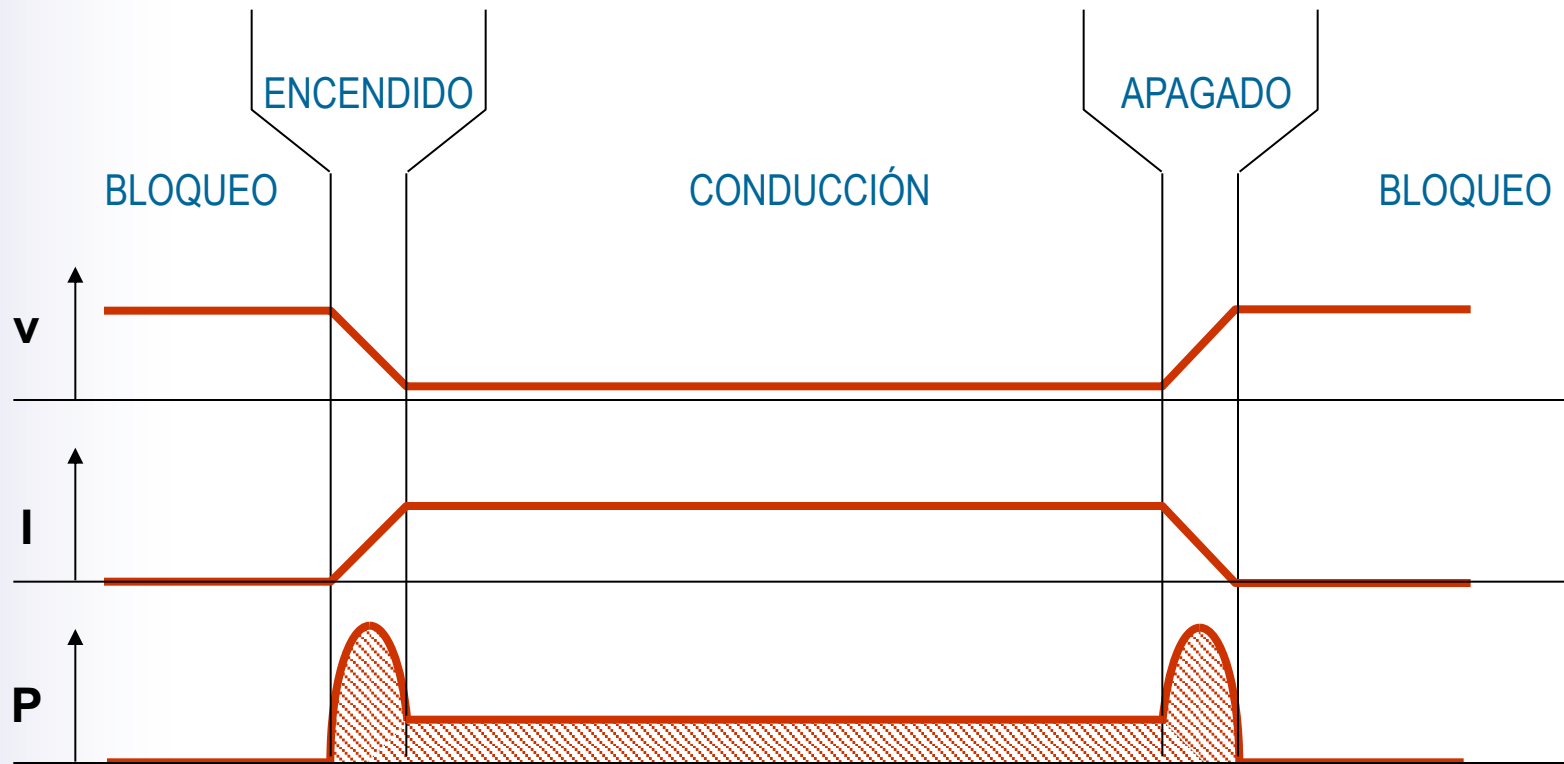


Power Switches



Dispositivos de potencia

Formas de onda típicas

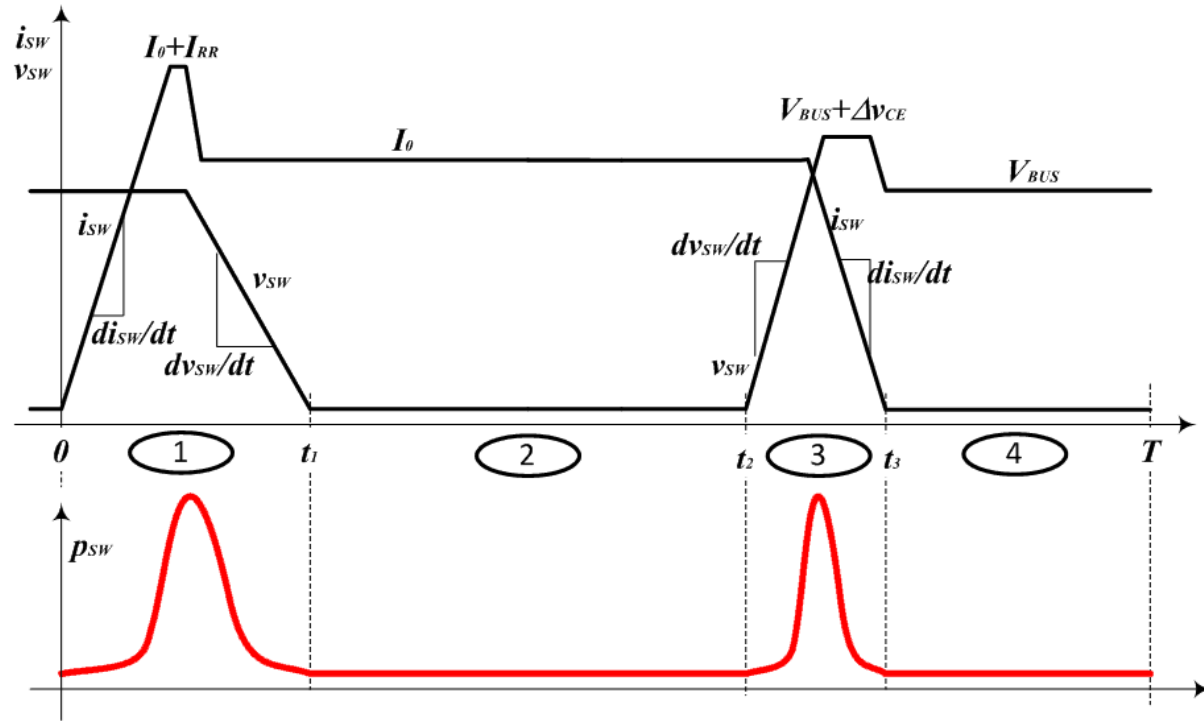
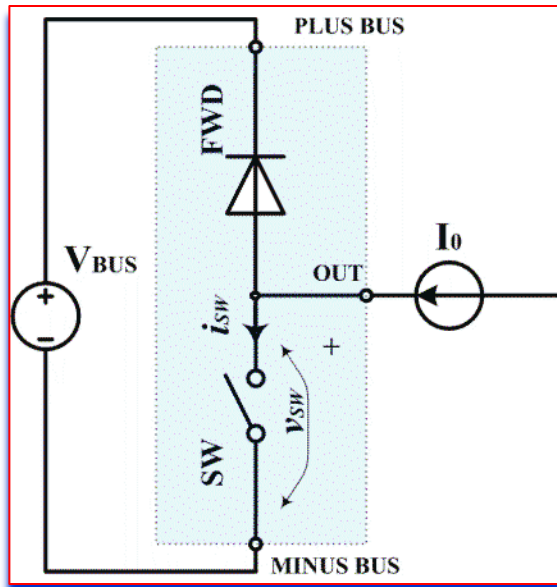


Los semiconductores actúan como interruptores casi ideales

Power Semiconductor Switch: Performances

Basic Switching Cell

- A core of any power converter



1. Turn-On switching losses
2. Conduction Losses
3. Turn-Off switching Losses
4. Blocking Losses

$$P = \frac{1}{T} \int_0^T v_{SW} i_{SW} dt = \frac{1}{T} \left(\underbrace{\int_0^{t_1} v_{SW} i_{SW} dt}_1 + \underbrace{\int_{t_1}^{t_2} v_{SW} i_{SW} dt}_2 + \underbrace{\int_{t_2}^{t_3} v_{SW} i_{SW} dt}_3 + \underbrace{\int_{t_3}^T v_{SW} i_{SW} dt}_4 \right)$$

Brief history of semiconductors

- 1874. Karl Ferdinand Braun → “unilateral conduction” of crystals
- 1894. Jagadish Chandra Bose → crystal for detecting radio waves
- 1925. Julius Edgar Lilienfeld → first patent on the Field Effect transistor
- 1940. Bell labs → inexpensive germanium diodes
- 1947. William Schokley and Gerald Pearson → first bipolar transistor
- 1959. Dawon Kahng and Martin M. (John) Atalla → first MOSFET
- 1979. Takashi Mimura → HEMT idea

2001 2008 2010 2010 2011 2012 2013 2016

Infienon
& Cree

SemiSouth

International
Rectifier

EPC

Cree

GaN
system

Fujitsu
Transphorm

Rohm

**SiC
diode**

**SiC
JFET**

**GaN
device**

eGaN

**SiC
MOSFET**

1200V GaN

600V GaN

**3rd generation
SiC MOSFET**

General Parameters to look at...

Physics

Static characteristics:

- ✓ ON: Voltage drop, i_{\max}
- ✓ OFF: Breakdown voltage

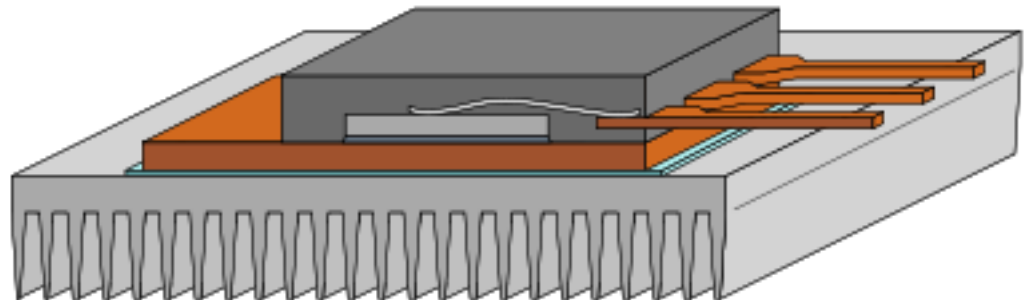
Dynamic characteristics :

- ✓ Rise & fall times, losses

Safe Operating Area

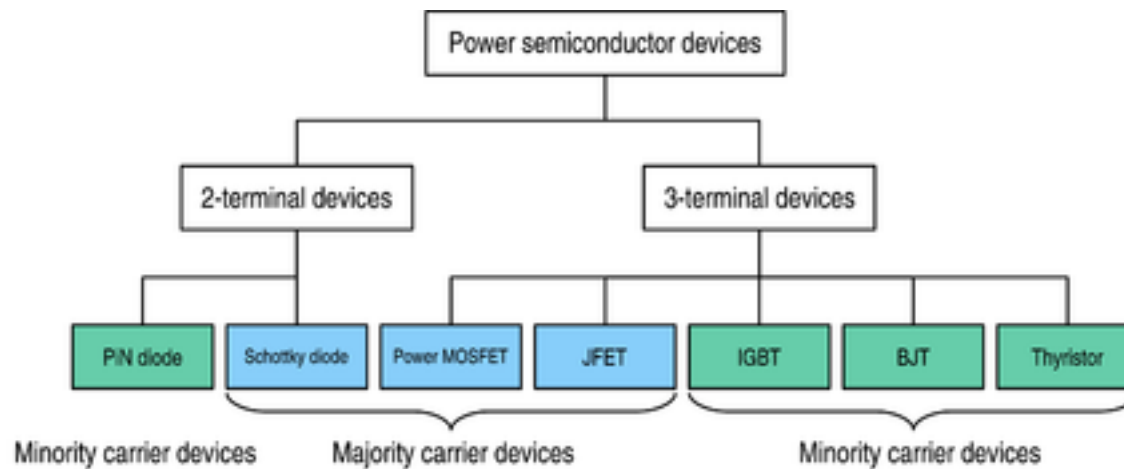
- ✓ Voltage, Current, Power, Latch-up

Package, Thermal resistance

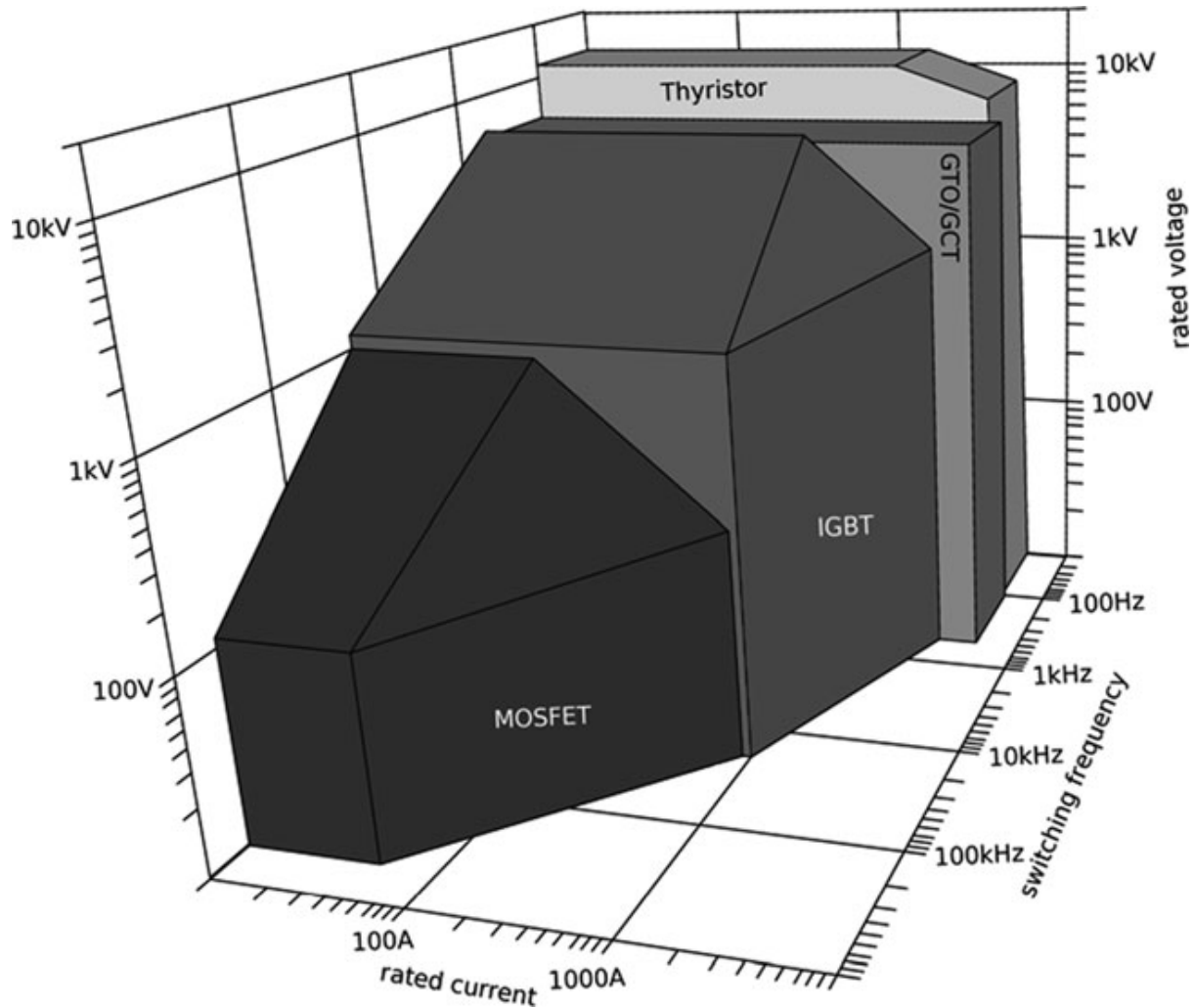




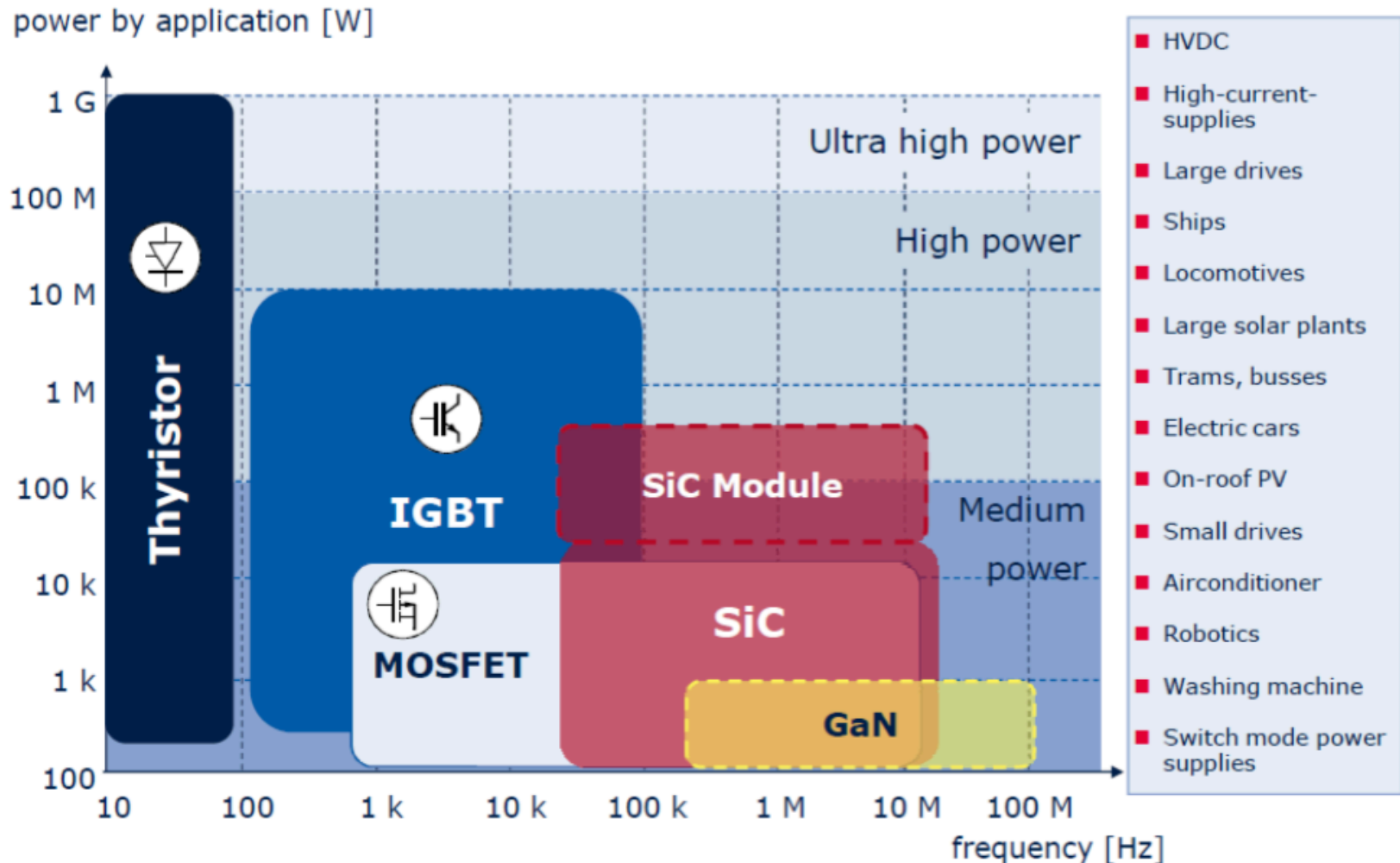
Classification, History & trends



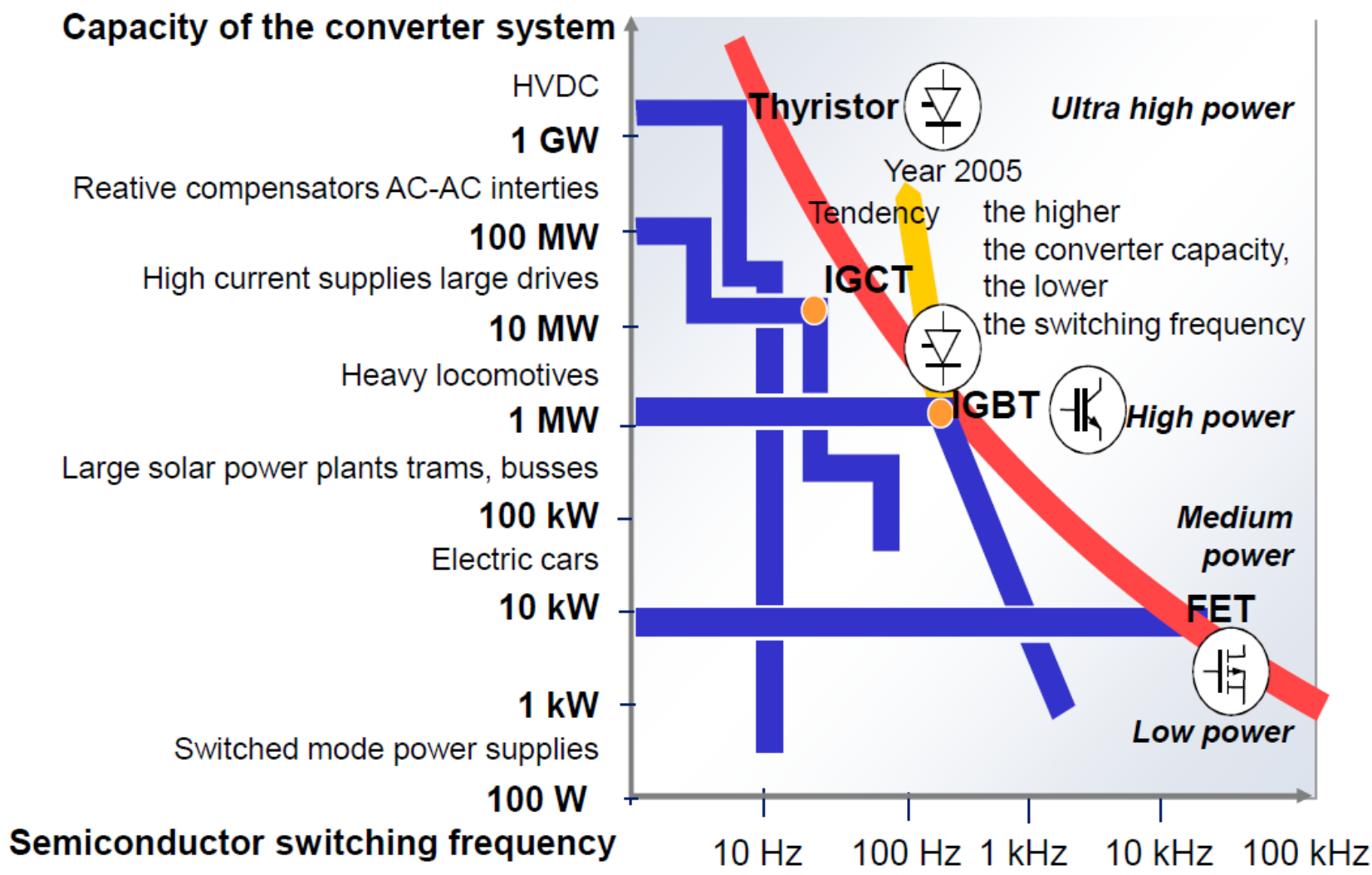
Power Semiconductor Switches



Infineon Portfolio of Power Technologies



Application for Power Semiconductor Components



Source: IPEC 2000

Power Semiconductor Switches

Diode	Thyristor	BJT	MOSFET	IGBT	
PiN	SCR	PNP	VMOSFET	PT	PLANAR
SBD (Schottky)	GTO	NPN	SuperJunction	NPT	TRENCH
	IGCT			SPT	
	Triac				

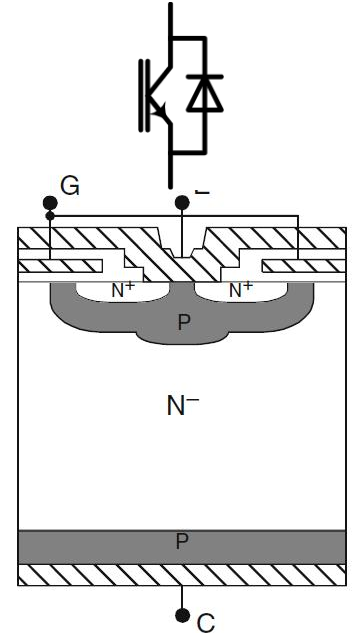
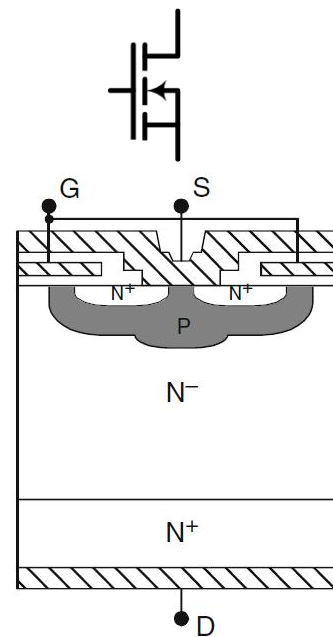
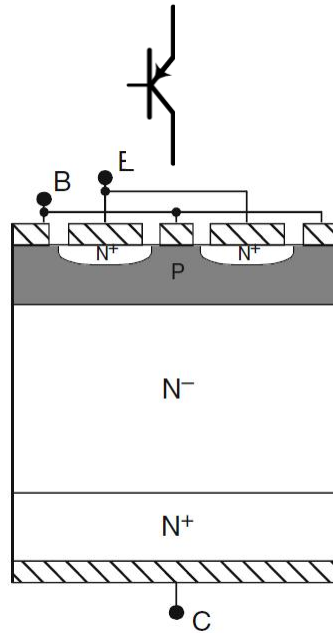
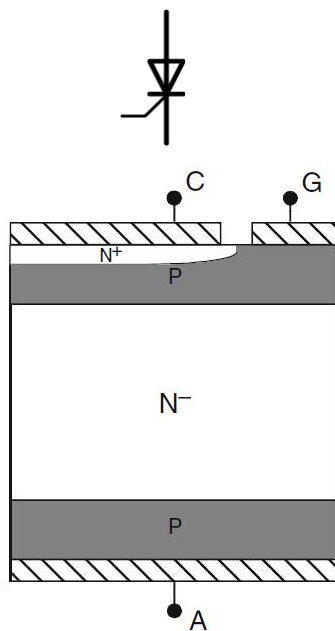
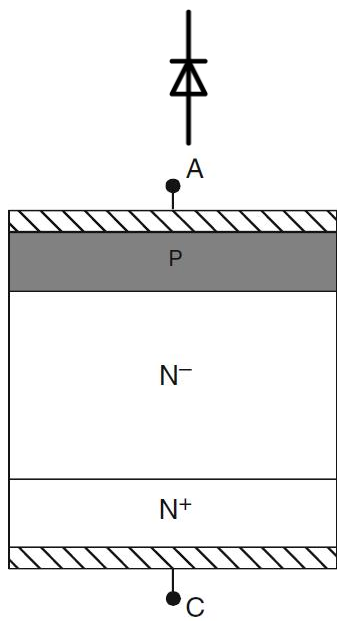


Table 1.2

Power electronics semiconductor devices

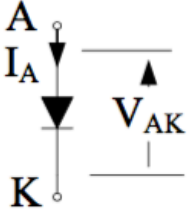
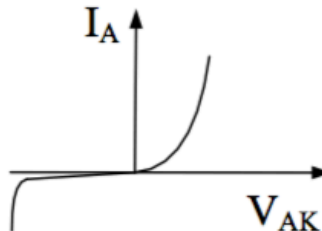
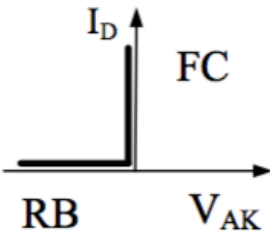
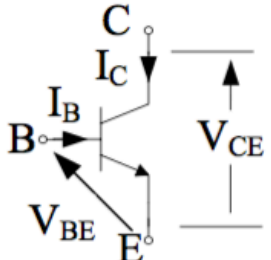
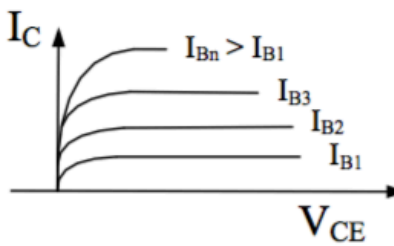
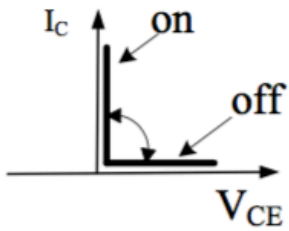
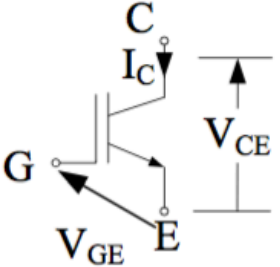
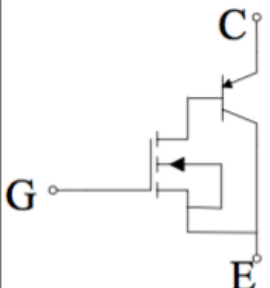
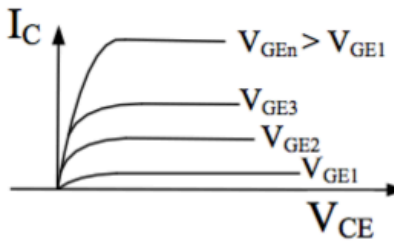
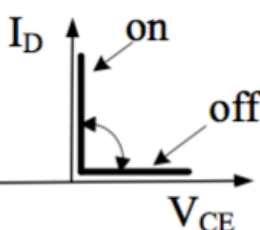
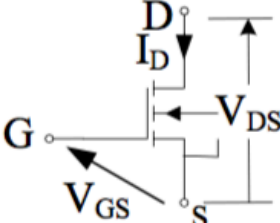
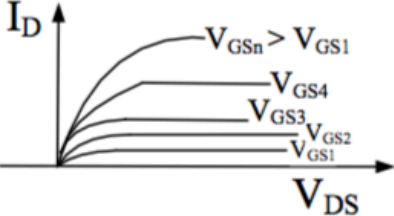
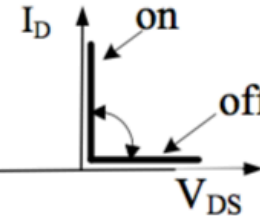
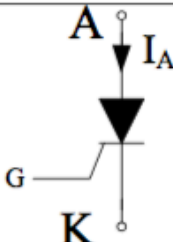
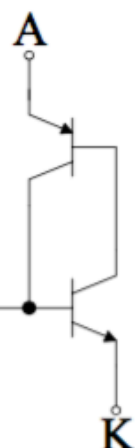
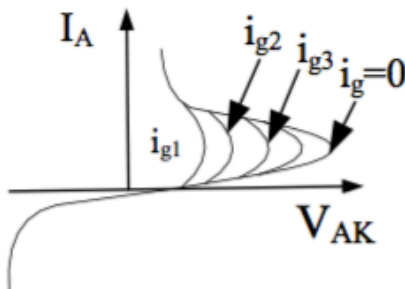
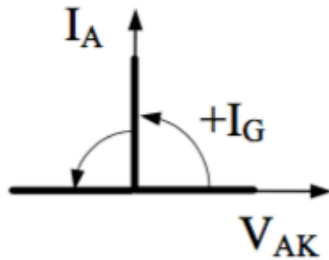
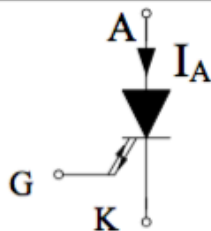
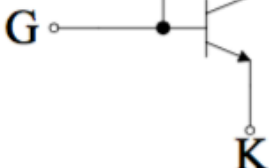
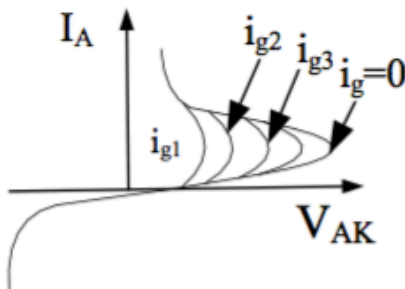
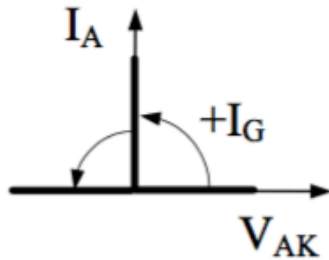
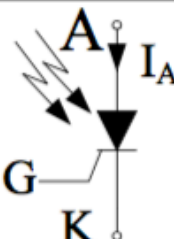
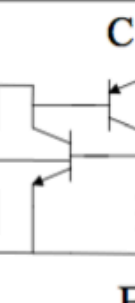
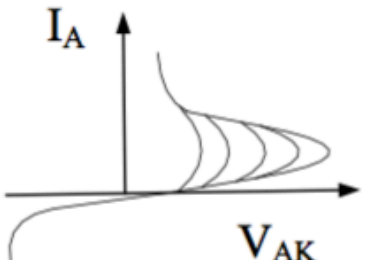
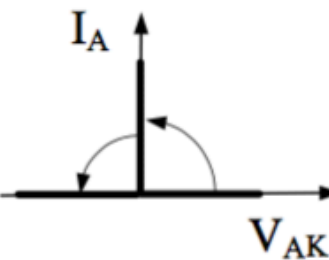
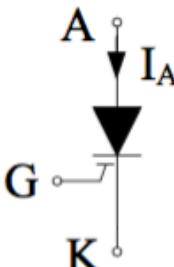
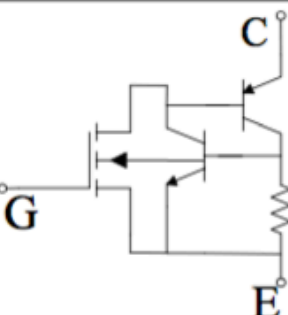
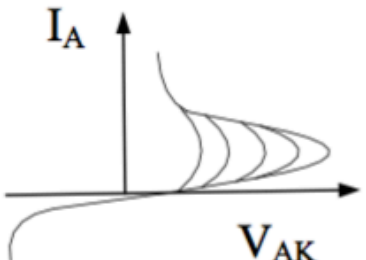
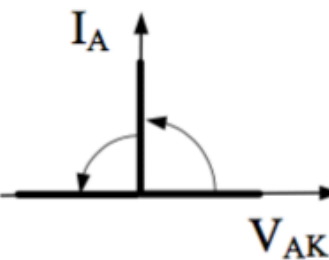
Power semiconductor device	Symbol	Equivalent circuit	I-V characteristics	Ideal I-V characteristics
Diode				
Power Transistor npn				
IGBT Insulated Gate Bipolar Transistor				
MOSFET				

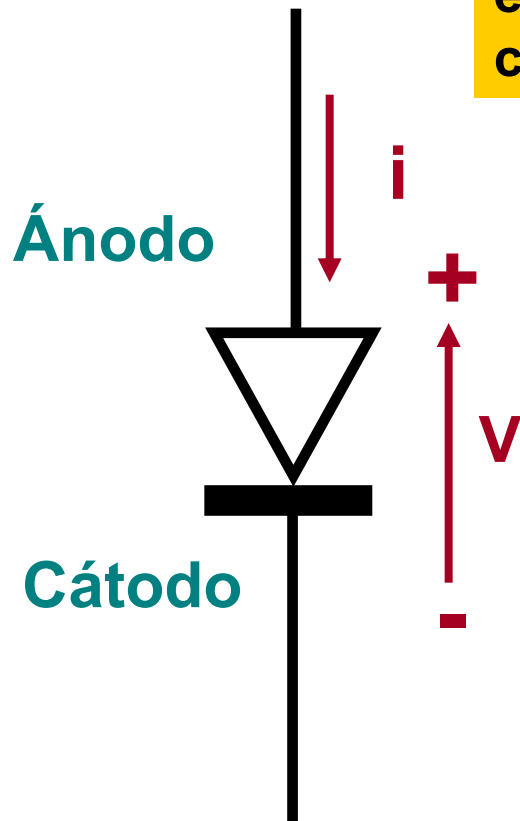
Table 1.2
Power electronics semiconductor devices

Power semiconductor device	Symbol	Equivalent circuit	I-V characteristics	Ideal I-V characteristics
Thyristor, SCR Silicon Control Rectifier			 $i_{g1} > i_{g2} > i_{g3} > \dots$	
GTO Gate Turn-off Thyristor			 $i_{g1} > i_{g2} > i_{g3} > \dots$	
LASCR Light Activated SCR Opto-Thyristor				
MCT MOS Controlled Thyristor				



Diodes

Concepto de diodo ideal



En polarización directa, la caída de tensión es nula, sea cual sea el valor de la corriente directa conducida

curva característica



En polarización inversa, la corriente conducida es nula, sea cual sea el valor de la tensión inversa aplicada

Ecuación característica del diodo

$$i = I_S \cdot (e^{\frac{V}{V_T}} - 1)$$

donde:

$$V_T = k \cdot T / q$$

$$I_S = A \cdot q \cdot n_i^2 \cdot (D_p / (N_D \cdot L_p) + D_n / (N_A \cdot L_n))$$

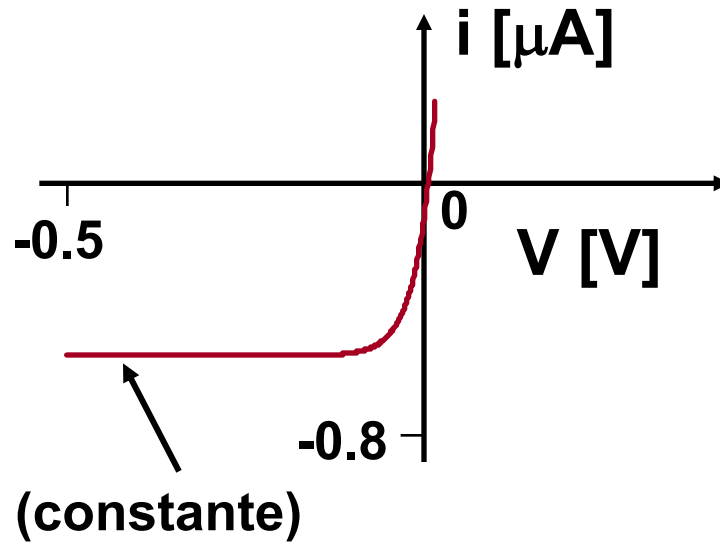
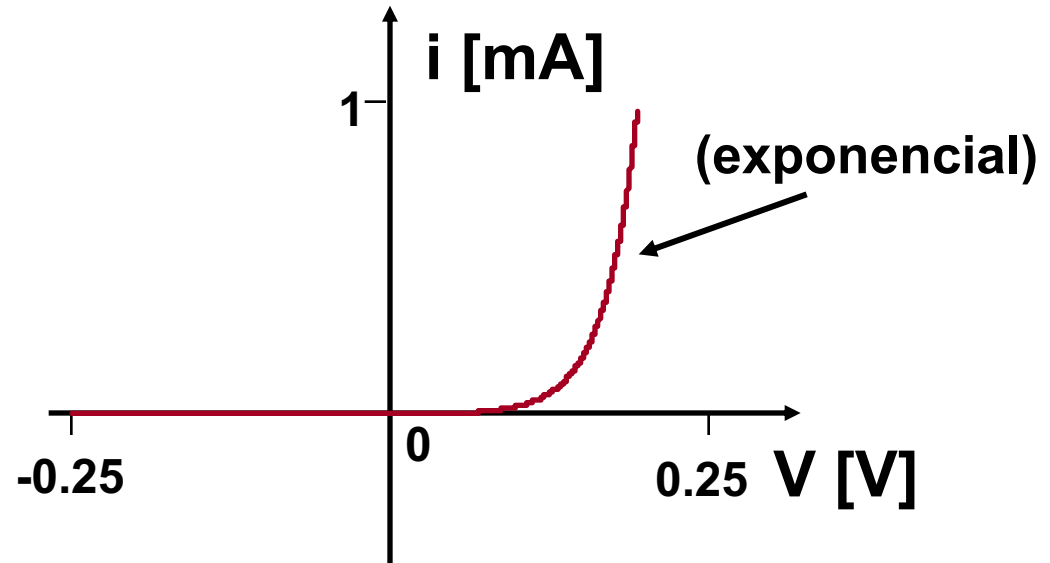
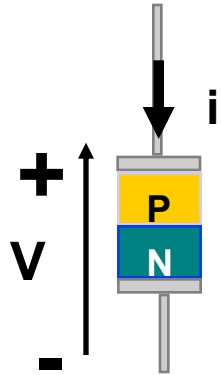
- *Polarización directa con $V_O > V \gg V_T$*

$$i \approx I_S \cdot e^{\frac{V}{V_T}} \quad (\text{dependencia exponencial})$$

- *Polarización inversa con $V \ll -V_T$*

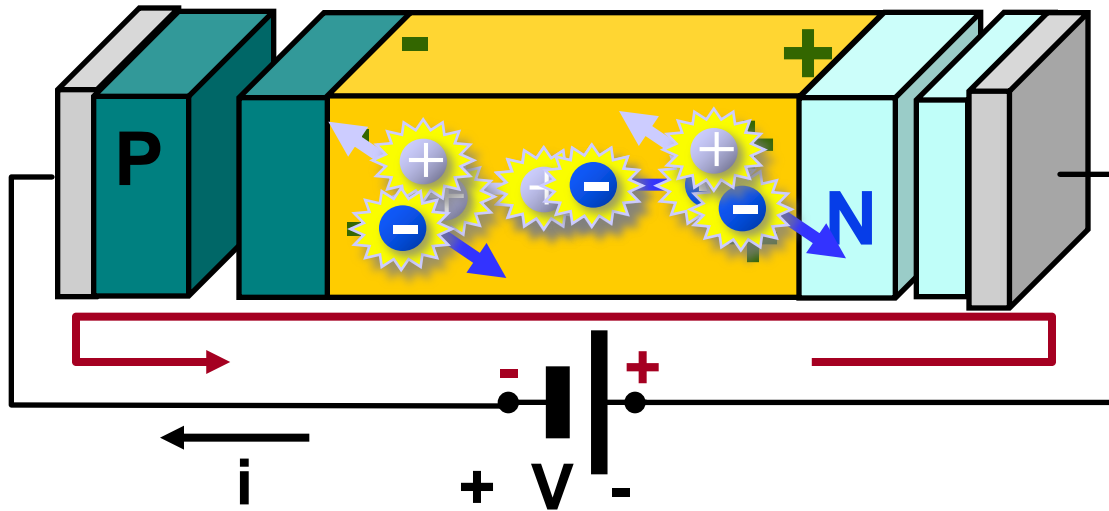
$$i \approx -I_S \quad \text{Corriente inversa de saturación (constante)}$$

Curva característica

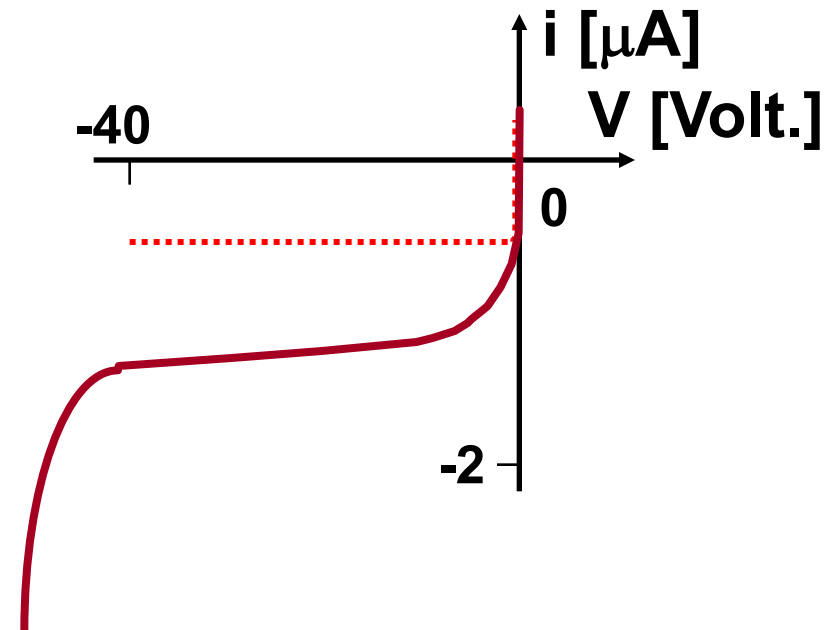


Curva característica

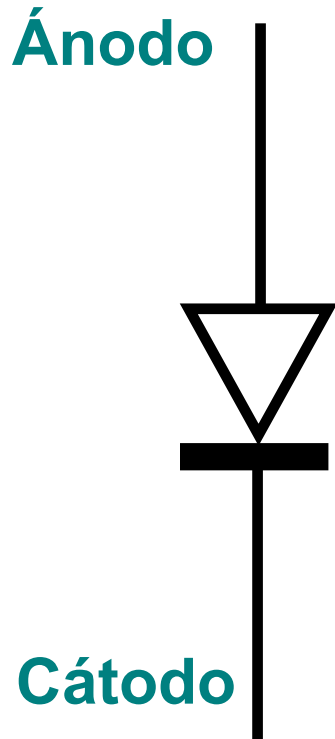
Avalancha primaria



La corriente aumenta fuertemente si se producen pares electrón-hueco adicionales.



El diodo semiconductor



Encapsulado
(cristal o resina
sintética)

Marca
señalando el
cátodo

Ánodo

Cátodo

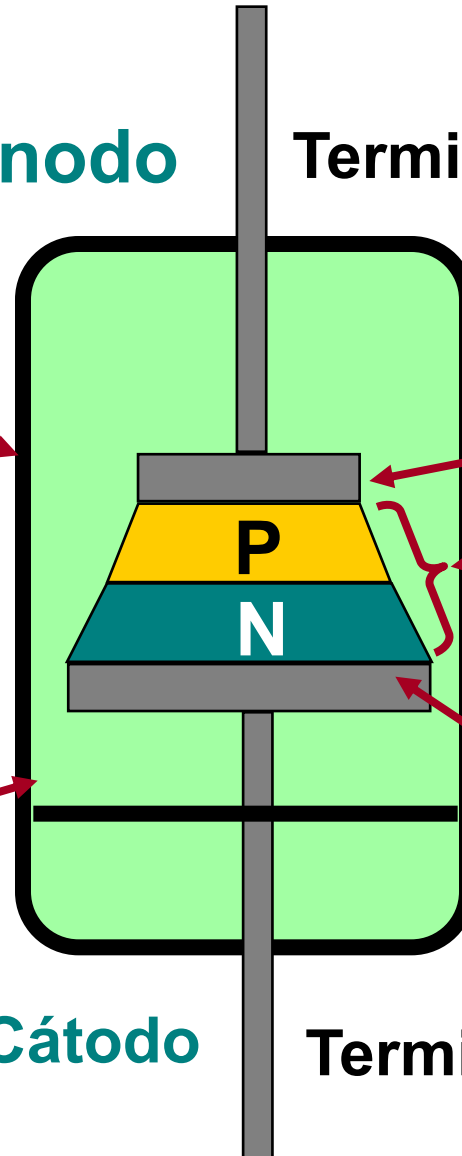
Terminal

Terminal

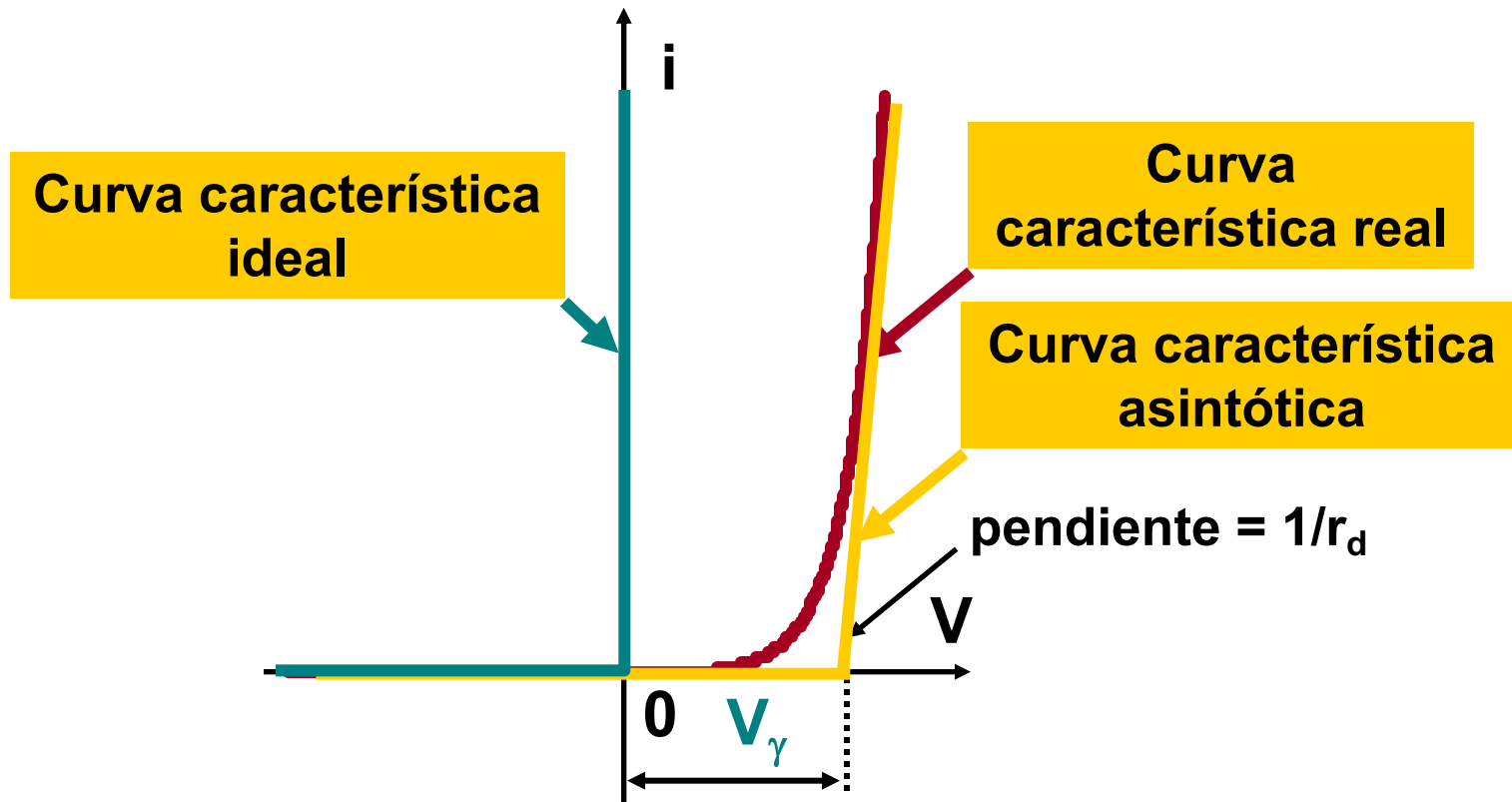
Contacto metal-
semiconductor

Oblea de
semiconductor

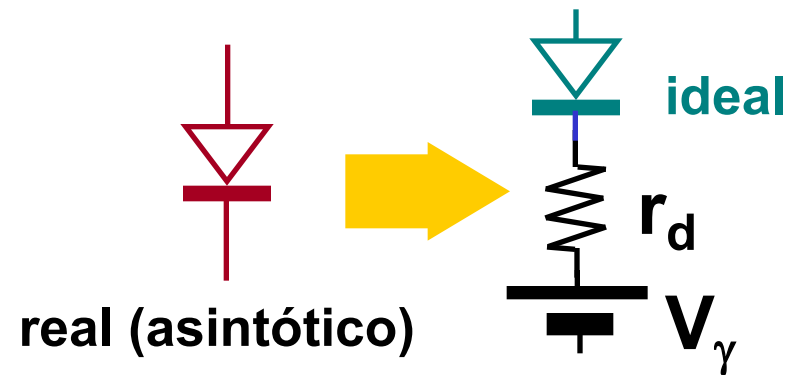
Contacto metal-
semiconductor



Curvas características y circuitos equivalentes



Circuito equivalente asintótico



Características fundamentales

- Tensión de ruptura
- Caída de tensión en conducción
- Corriente máxima
- Velocidad de conmutación

Tensión de ruptura

Baja tensión

15 V

30 V

45 V

55 V

60 V

80 V

Media tensión

100 V

150 V

200 V

400 V

Alta tensión

500 V

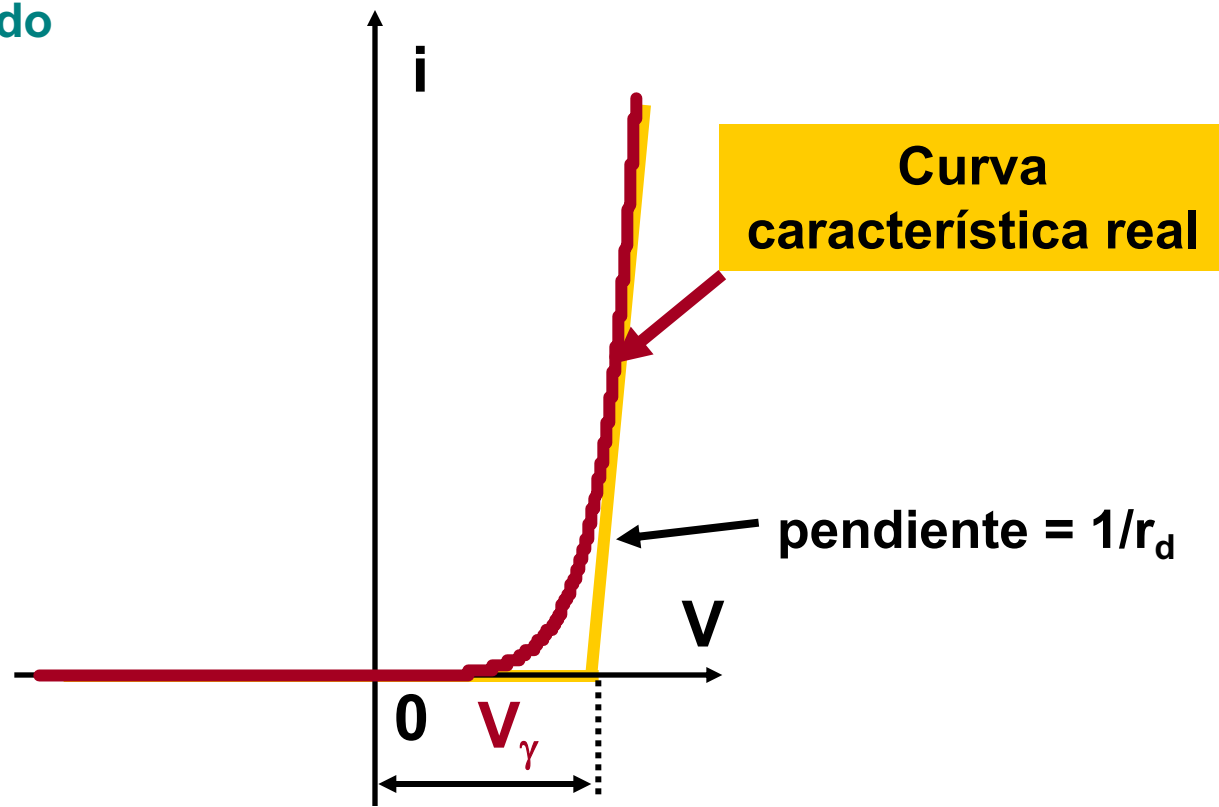
600 V

800 V

1000 V

1200 V

Tensión de codo

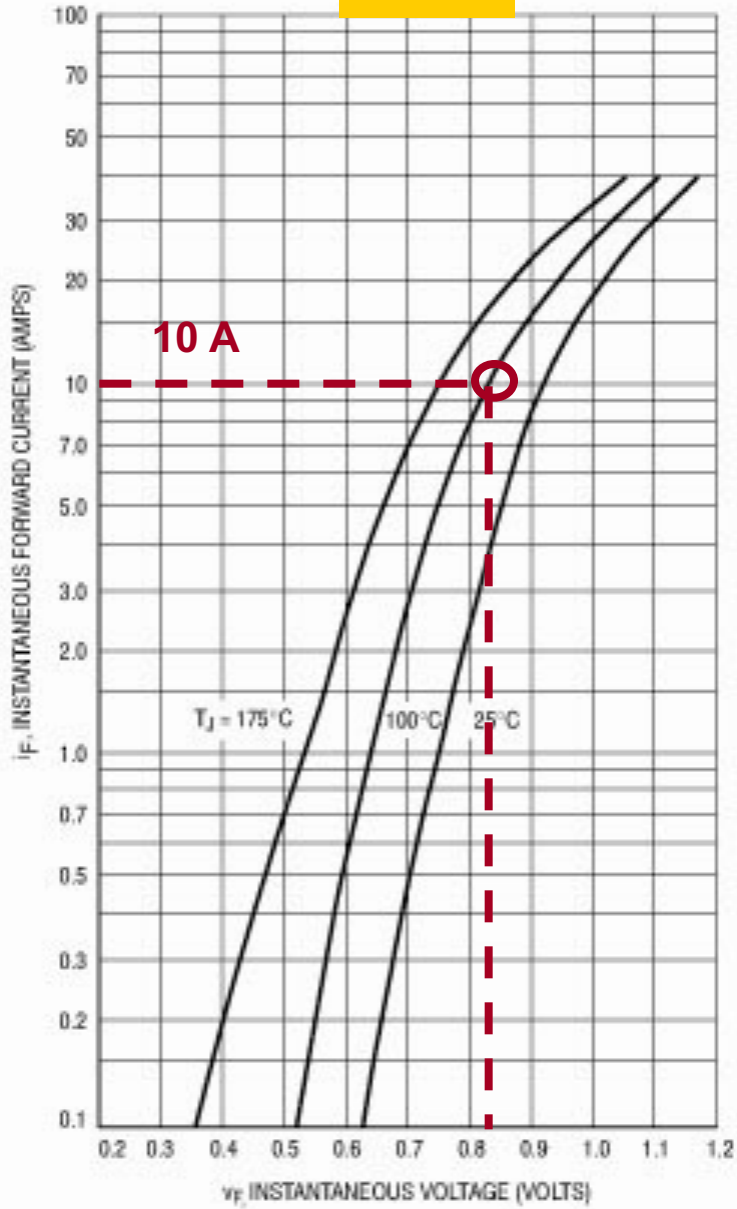


A mayor tensión de ruptura , mayor caída de tensión en conducción

	Señal	Potencia	Alta tensión
V_{Ruptura}	$< 100 \text{ V}$	$200 - 1000 \text{ V}$	$10 - 20 \text{ kV}$
V_{Codo}	$0,7 \text{ V}$	$< 2 \text{ V}$	$> 8 \text{ V}$

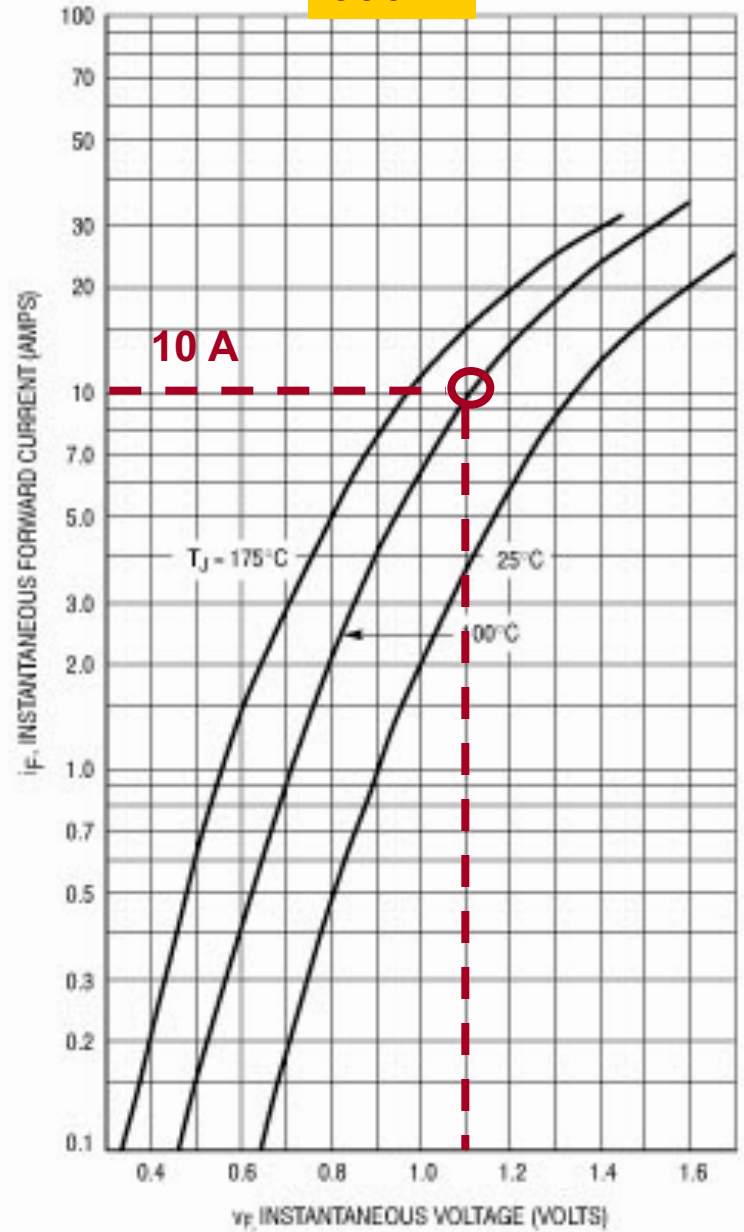
Curva real de un dispositivo

200 V



0,82 V

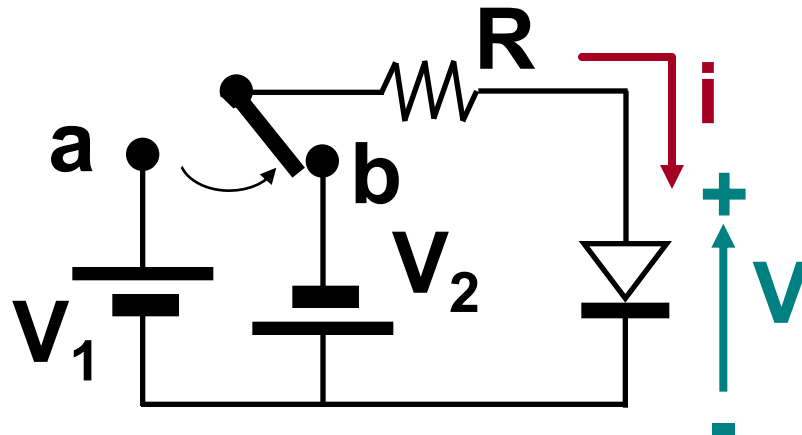
600 V



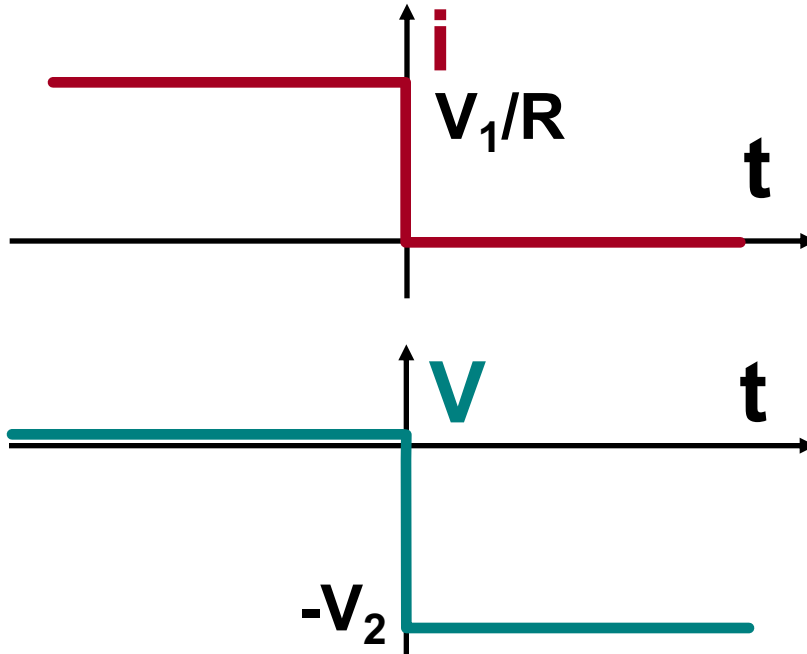
1,1 V

Características dinámicas

Indican capacidad de conmutación del diodo



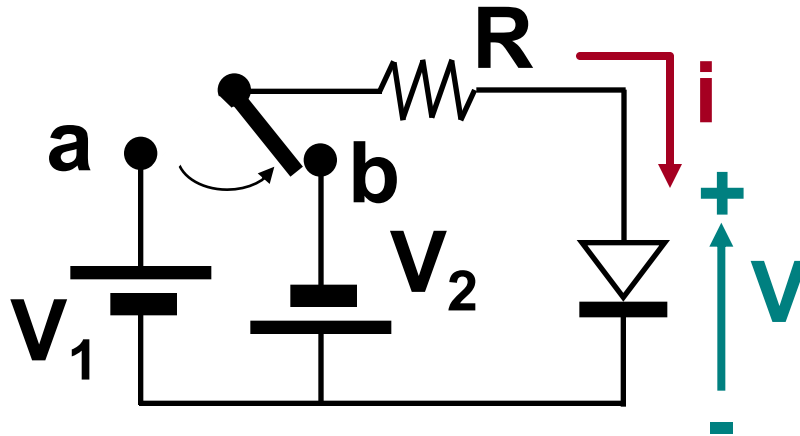
Transición de “a” a “b”



Comportamiento
dinámicamente ideal

Características dinámicas

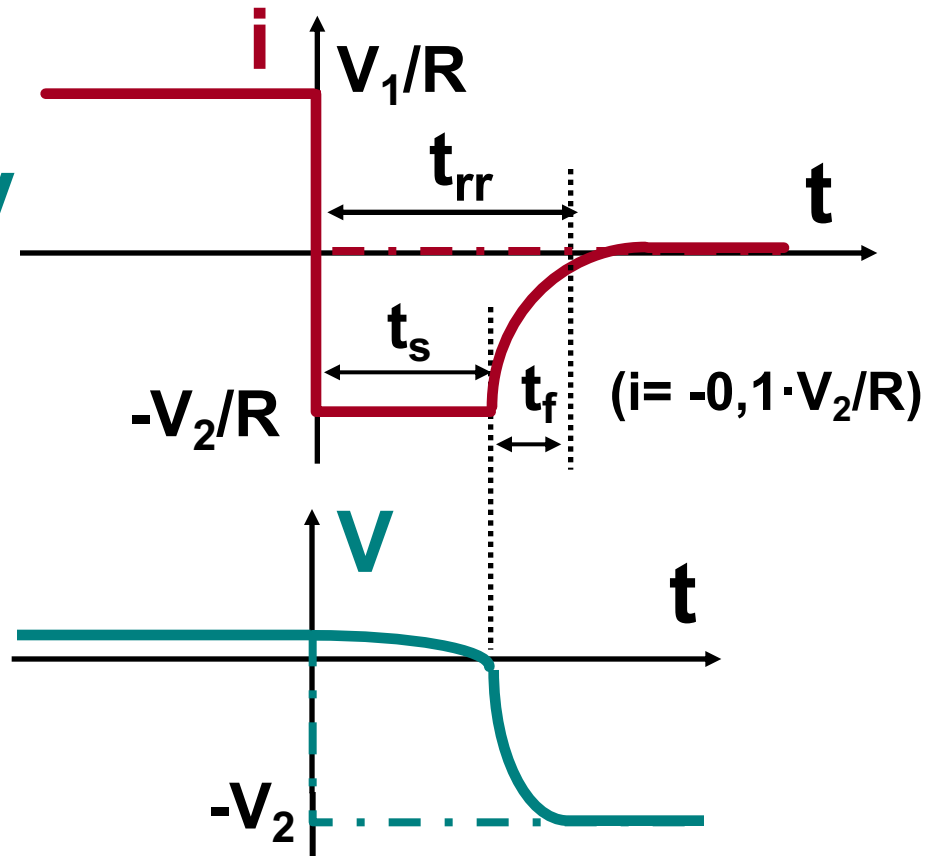
Transición de “a” a “b”



t_s = tiempo de almacenamiento
(storage time)

t_f = tiempo de caída (fall time)

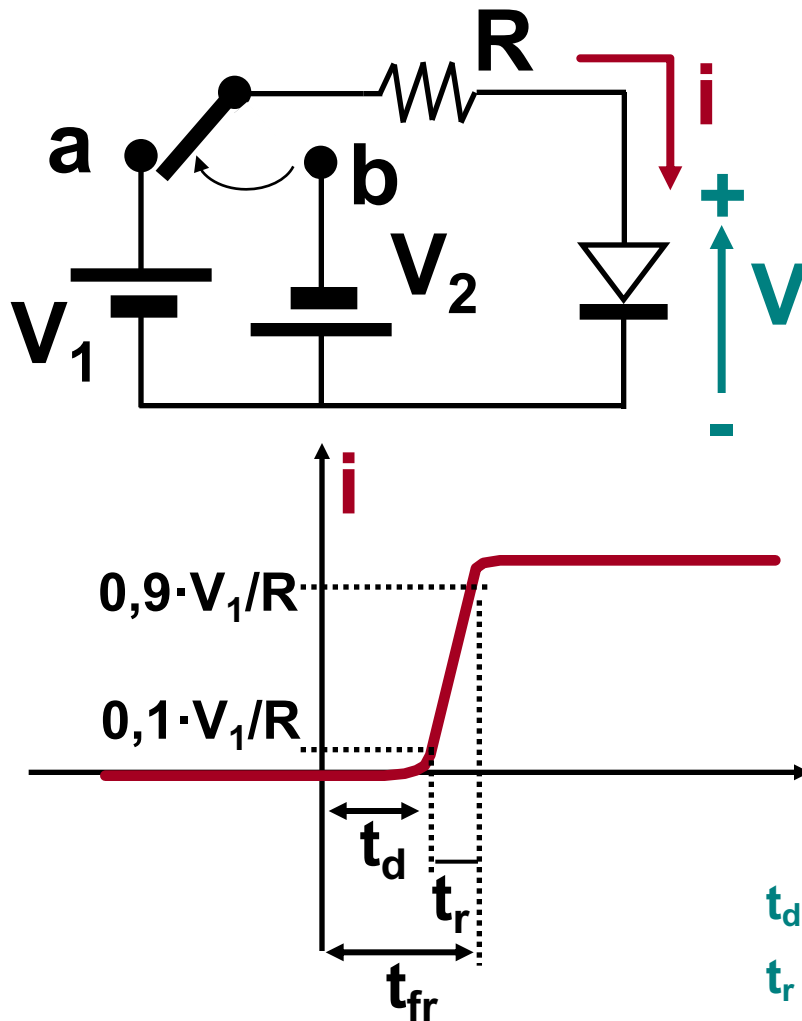
t_{rr} = tiempo de recuperación
inversa (reverse recovery time)



Características dinámicas

Transición de "b" a "a" (encendido)

El proceso de encendido es más rápido que el apagado.



t_d = tiempo de retraso (delay time)

t_r = tiempo de subida (rise time)

$t_{fr} = t_d + t_r$ = tiempo de recuperación directa (forward recovery time)

Características Principales

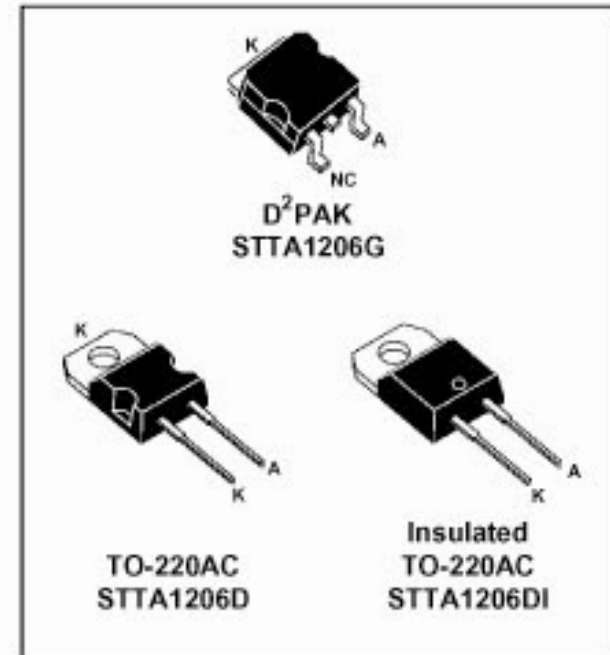
Corriente directa
Tensión inversa
Tiempo de recuperación
Caída de tensión
en conducción

MAIN PRODUCT CHARACTERISTICS

$I_F(AV)$	12A
V_{RRM}	600V
$t_{rr} (typ)$	28ns
$V_F (max)$	1.5V

FEATURES AND BENEFITS

- SPECIFIC TO "FREEWHEEL MODE" OPERATIONS: FREEWHEEL OR BOOSTER DIODE.
- ULTRA-FAST AND SOFT RECOVERY.
- VERY LOW OVERALL POWER LOSSES IN BOTH THE DIODE AND THE COMPANION TRANSISTOR.
- HIGH FREQUENCY OPERATIONS.
- INSULATED PACKAGE : TO-220AC
Electrical insulation : 2500V_{RMS}
Capacitance < 7 pF



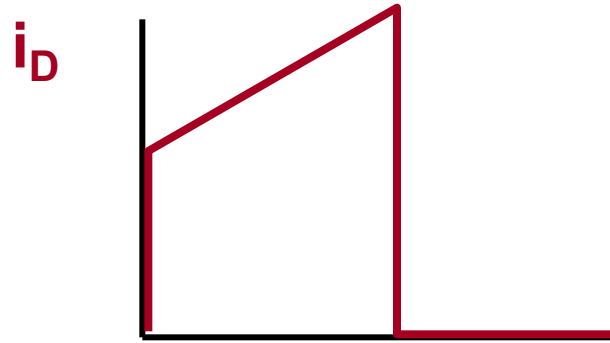
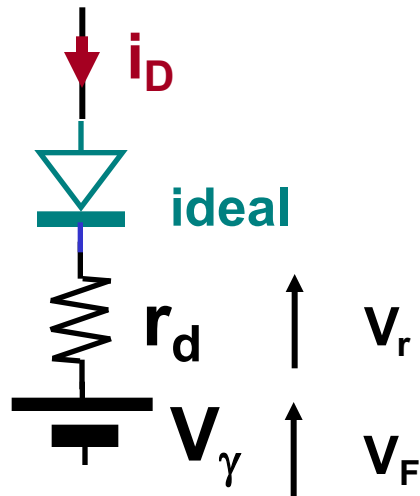
Encapsulado

Pérdidas en el diodo

2 Tipos

- Estáticas
- Dinámicas

Pérdidas estáticas



$$P_D(t) = v_D(t) \cdot i_D(t) = (V_\gamma + r_D \cdot i(t)) \cdot i(t)$$

$$P_D = \frac{1}{T} \int_0^T P_D(t) \cdot dt$$

$$P_D = V_\gamma \cdot I_D + r_D \cdot I_{ef}^2$$

I_D : Valor medio

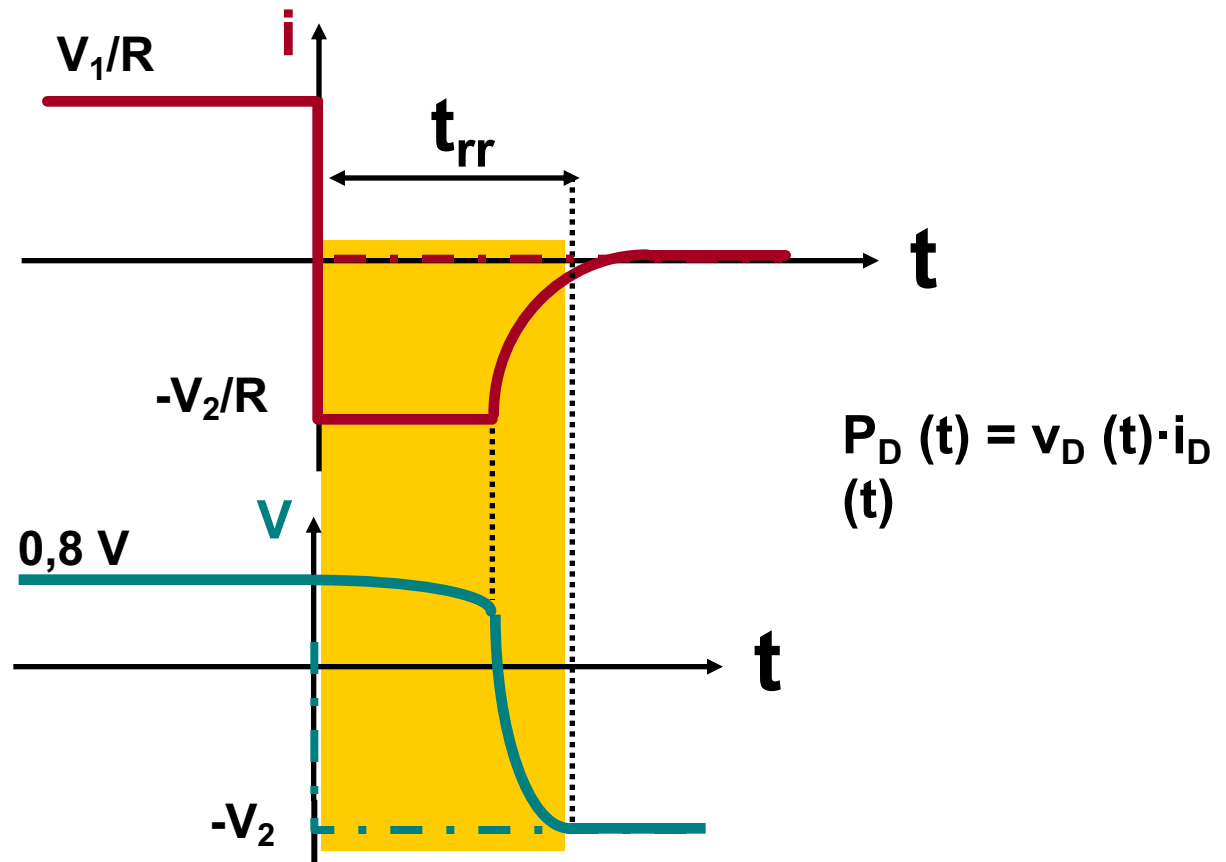
I_{ef} : Valor eficaz

Pérdidas en el diodo

Pérdidas dinámicas (*Perdidas de conmutación*)

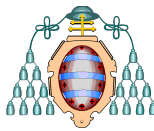
Las conmutaciones no son perfectas

Hay instantes en los que conviven tensión y corriente

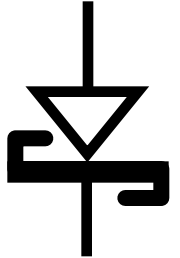




Rectifying contacts (III)



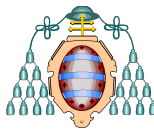
- There is a type of diode based on the operation of a rectifying contact. It is the **Schottky diode**. Schottky diodes are widely used in many applications, including RF (telecom) circuits and low-voltage, medium-power power converters.
- Main features:
 - Lower forward voltage drop than a similar-range, PN-junction diode.
 - They are faster than PN diodes because minority carriers hardly play any role in the current conduction process. They are **majority carrier devices**.
 - However, they always have a higher reverse current (this is not a big problem).
 - When they are made up of silicon, the maximum reverse voltage (compatible with reasonable drop voltage in forward bias) is about 200 V.
 - However, Schottky diodes made up of wide-bandgap materials (such as silicon carbide and gallium nitride) reach breakdown voltages as high as 600-1200 V.



Schematic
Symbol

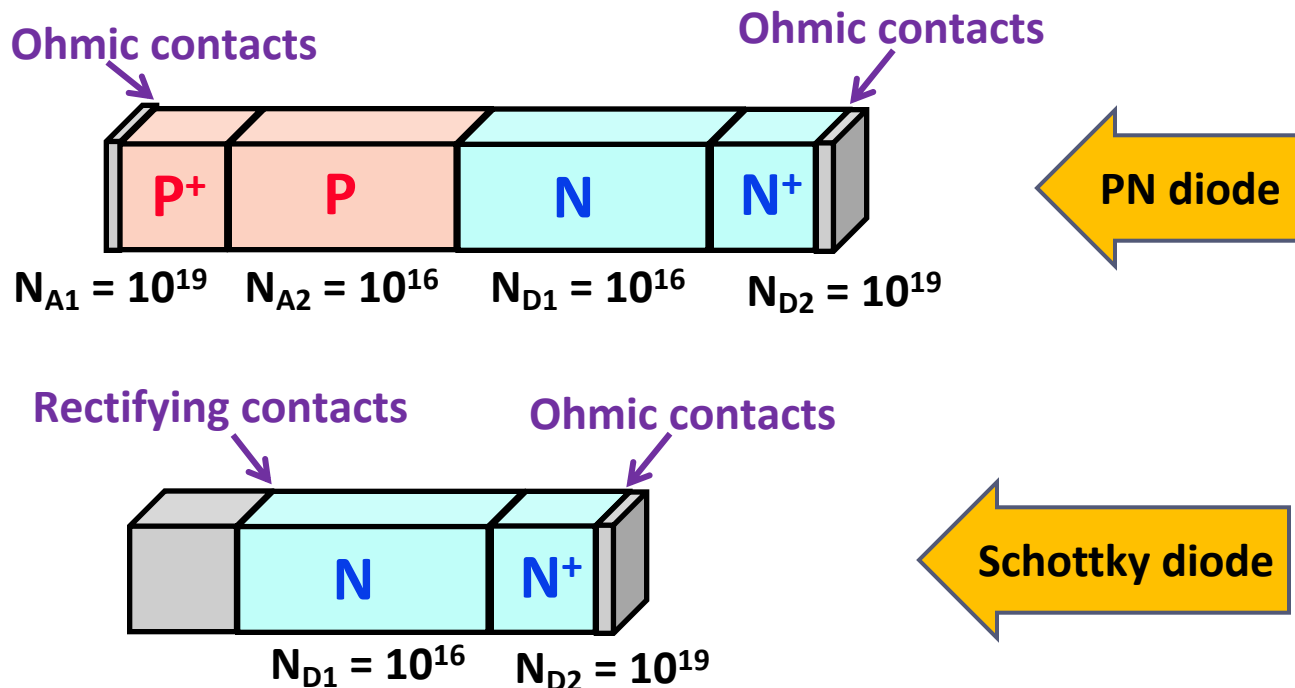


Ohmic contacts



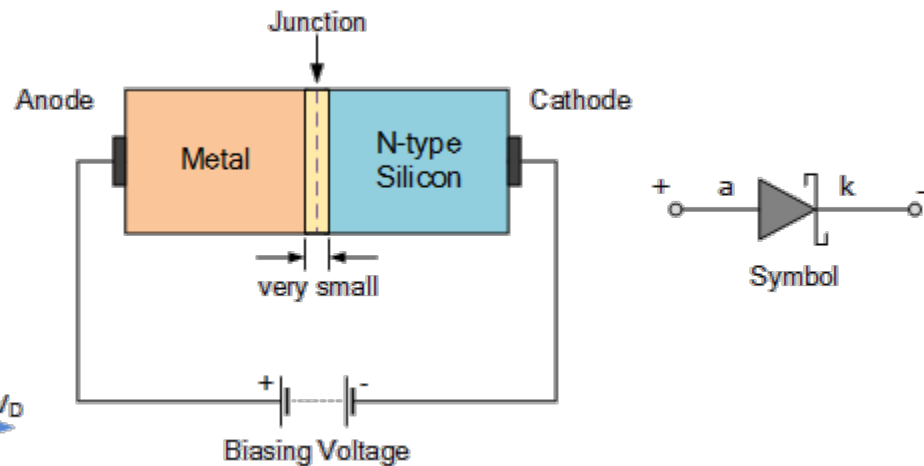
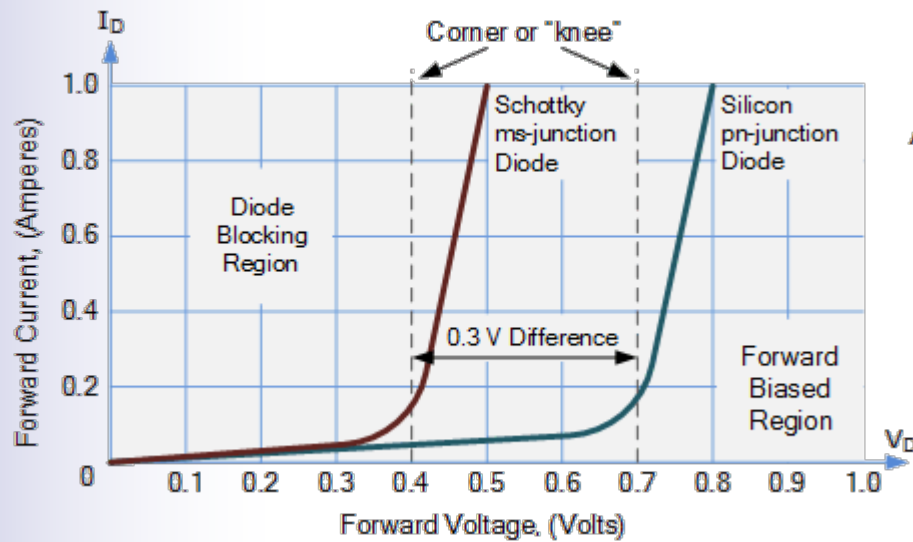
- There are two different possibilities to obtain ohmic contacts:
 - a) According to the previous slides, to have an MS junction corresponding to Case #3 or to Case #4.
 - b) To have MS junctions corresponding to Cases #1 or #2, but with an extremely-doped semiconductor side (10^{19} atoms/cm³). In this situation, electrons can flow in both directions by a *tunneling* process.

Beyond the scope of this course, as well.



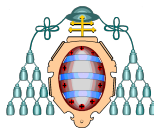


Schotky diode

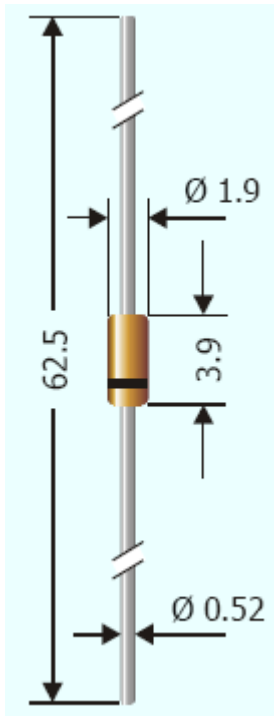




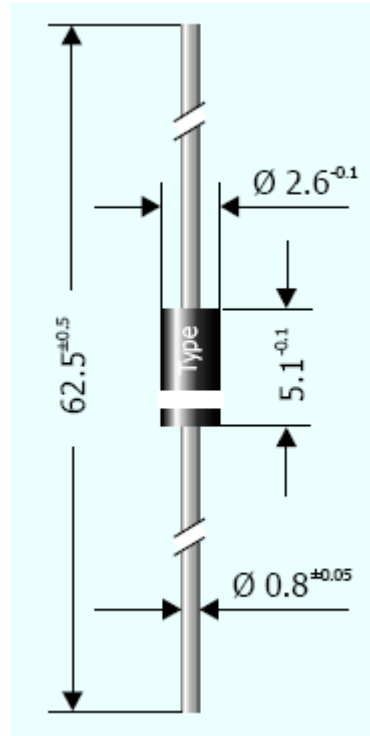
Packages for diodes (I)



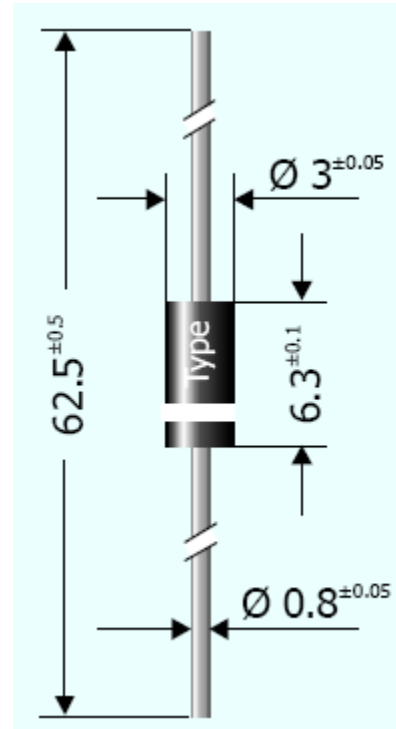
- Axial leaded through-hole packages (low power).



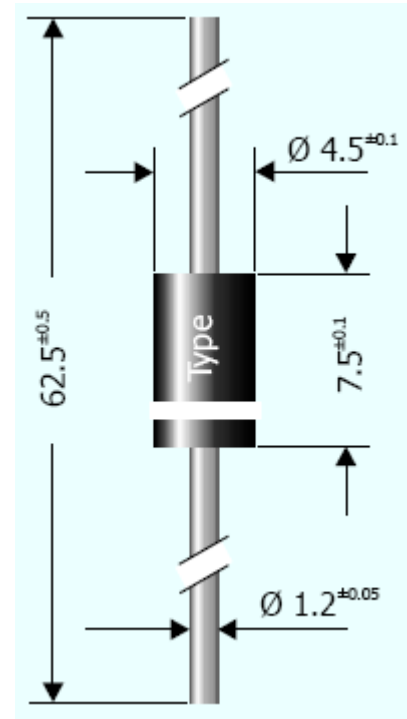
DO 35



DO 41



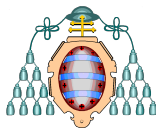
DO 15



DO 201

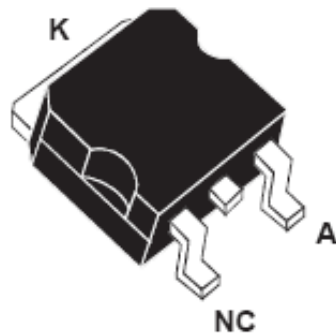
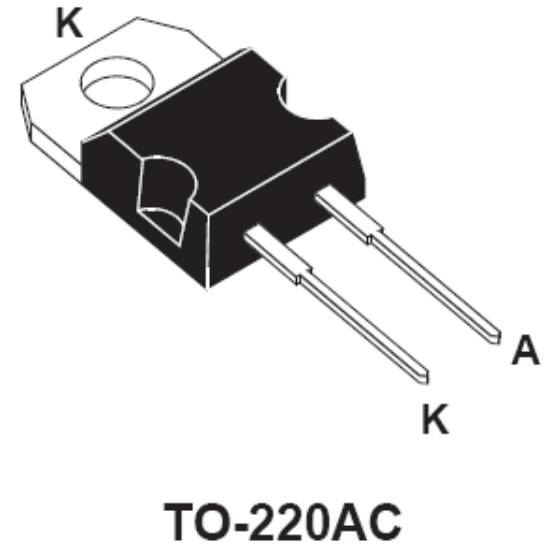
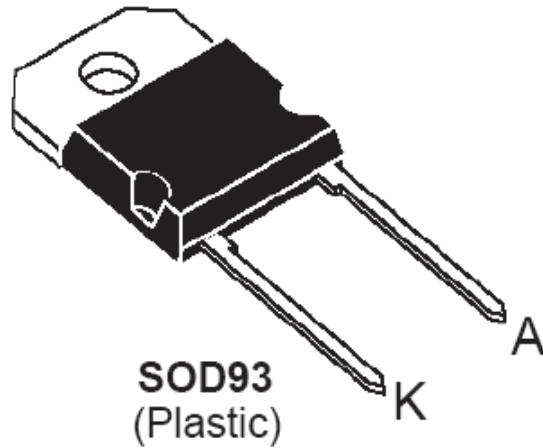


Packages for diodes (II)

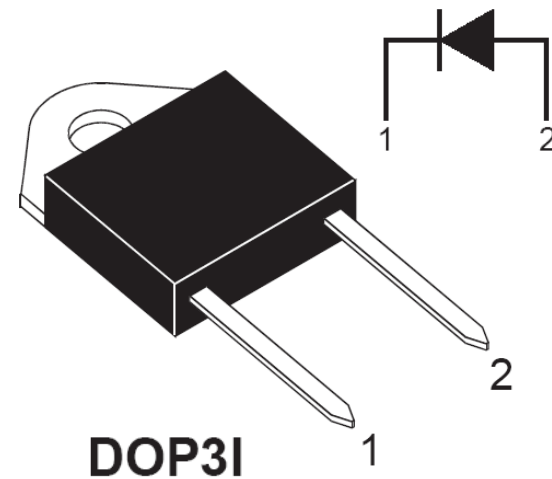


- Packages to be used with heat sinks.

Cathode connected to case



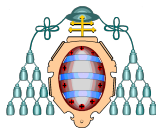
D²PAK



DOP31



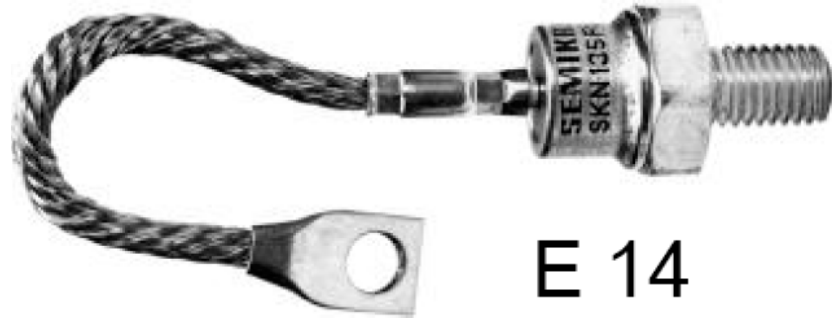
Packages for diodes (III)



- Packages to be used with heat sinks (higher power levels).



DO 5



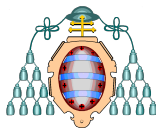
E 14



B 44

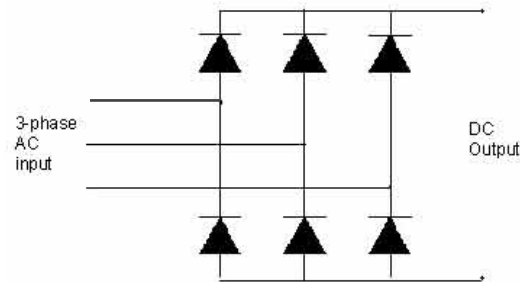


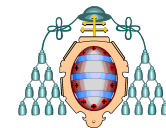
E 35



Packages for diodes (XI)

- Assembly of 6 diodes
(Three-phase bridge rectifiers)

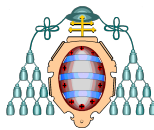




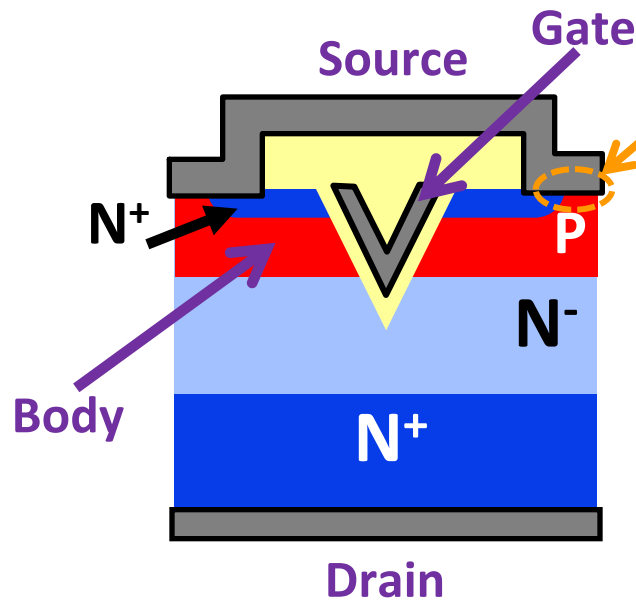
MOSFET



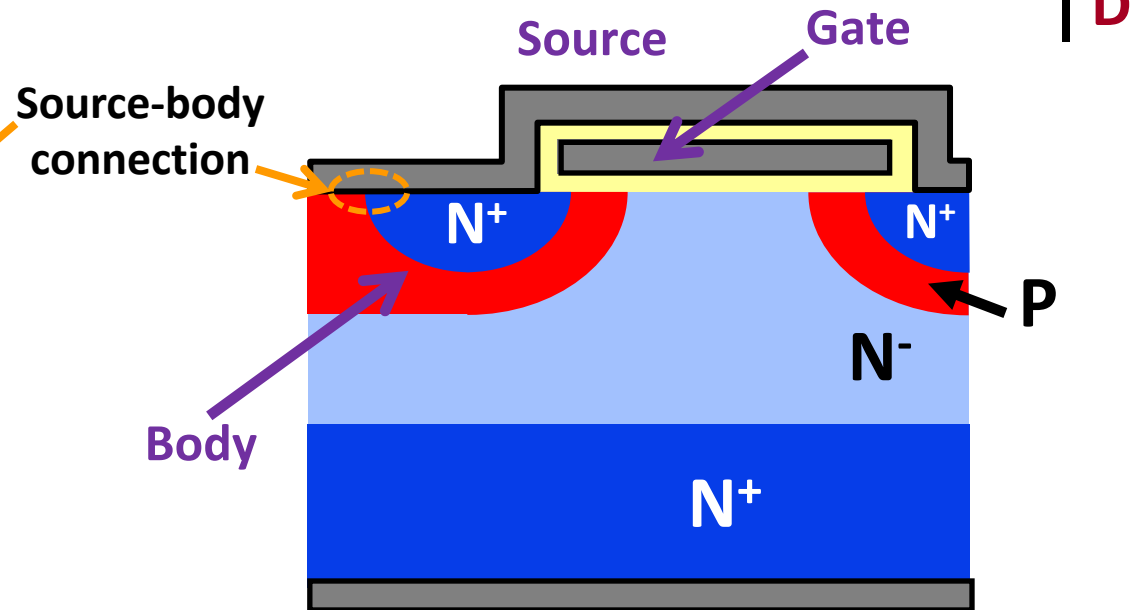
Internal structure of power MOSFETs (I).



- A typical transistor is constituted of several thousand cells.
- As all the FET devices, MOSFET can be easily connected in parallel.
- They are vertical MOSFETs.
- Examples of cells:



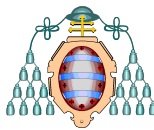
V-groove MOS (VMOS)



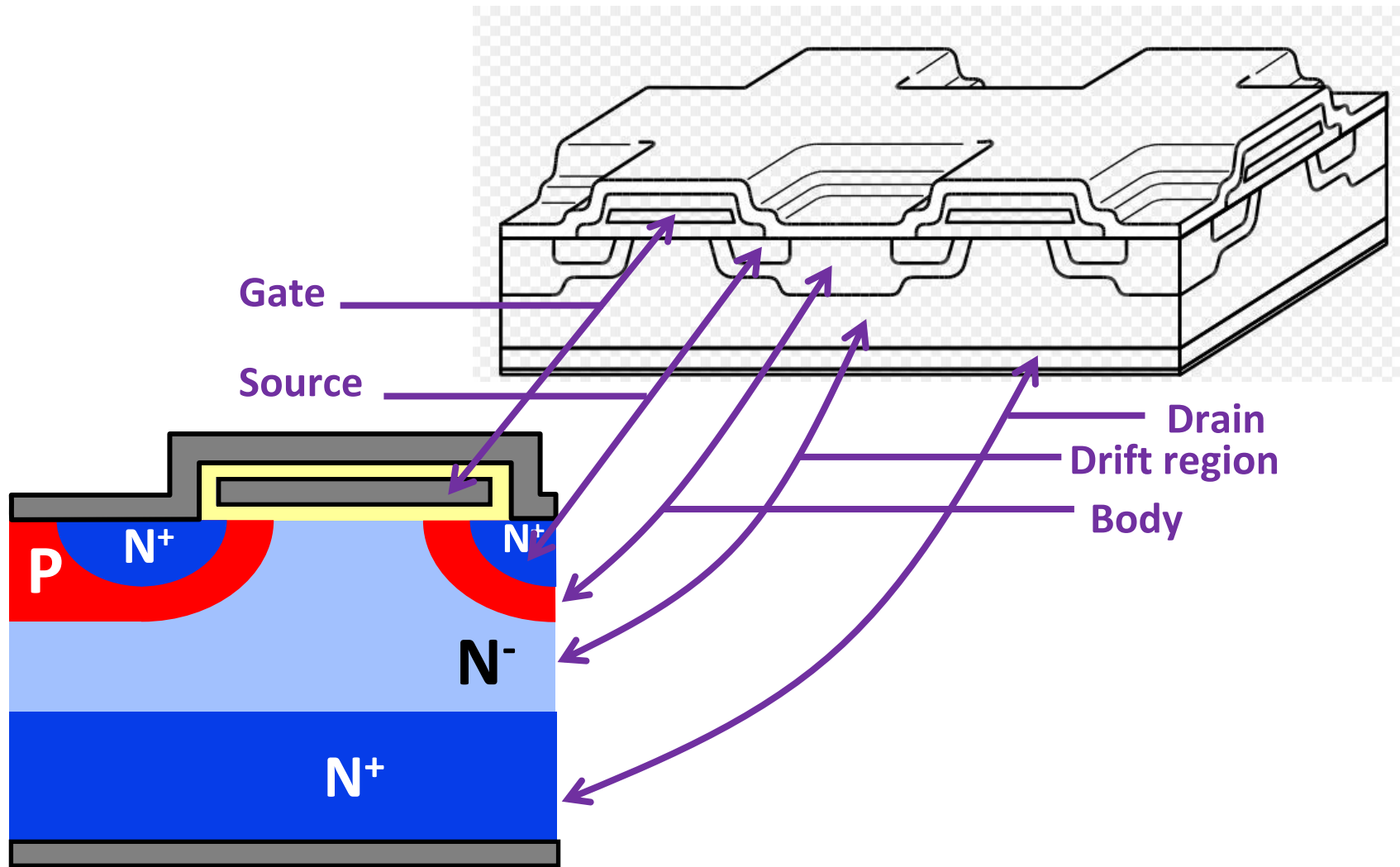
Double diffused MOSFET (DMOS)



Internal structure of power MOSFETs (IV).

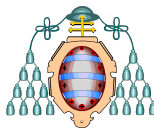


- **Tridimensional structure of a DMOS:**

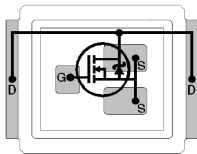




Packages for power MOSFETs (I).



- Packages are similar to those of power diodes (except axial leaded through-hole packages).
- There are many different packages.
- Examples: 60V MOSFETs.



DirectFET™ ISOMETRIC

IRF6648

$$R_{DS(on)} = 5.5 \text{ m}\Omega, I_D = 86 \text{ A}$$



SO-8

IRF7855 **PbF**

$$R_{DS(on)} = 9.4 \text{ m}\Omega, I_D = 12 \text{ A}$$



TO-220AB

IRFZ44VZ **PbF**



D²Pak

IRFZ44VZS **PbF**



TO-262

IRFZ44VZL **PbF**

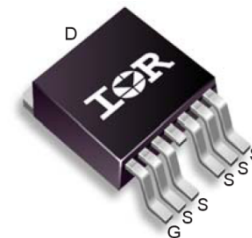
$$R_{DS(on)} = 12 \text{ m}\Omega, I_D = 57 \text{ A}$$



TO-247AC

IRFP054V **PbF**

$$R_{DS(on)} = 9 \text{ m}\Omega, I_D = 93 \text{ A}$$



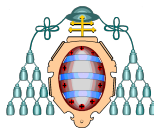
D²Pak 7 Pin

IRFS3006-7P **PbF**

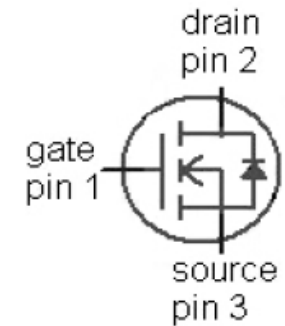
$$R_{DS(on)} = 1.5 \text{ m}\Omega, I_D = 240 \text{ A}$$



Packages for power MOSFETs (II).



- Other examples of 60V MOSFETs.



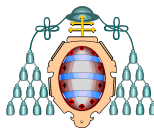
Type	IPB034N06L3 G	IPI037N06L3 G	IPP037N06L3 G
Package	PG-TO-263-3	PG-TO-262-3	PG-TO-220-3
Marking	034N06L	037N06L	037N06L

$$R_{DS(on)} = 3.4 \text{ m}\Omega, I_D = 90 \text{ A}$$

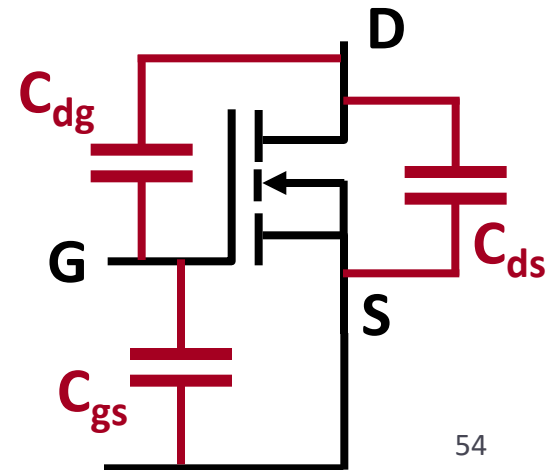




Parasitic capacitances in power MOSFETs (I).

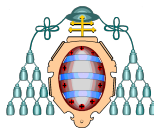


- Power MOSFETs are faster than other power devices (such as bipolar transistors, IGBTs, thyristors, etc.).
- This is because MOSFETs are unipolar devices (no minority carriers are stored at the edges of PN junctions).
- The switching speed is limited by parasitic capacitances.
- Three parasitic capacitances should be taken into account:
 - C_{gs} , which is a quite linear capacitance.
 - C_{ds} , which is a non-linear capacitance.
 - C_{dg} , Miller capacitance, which is also a non-linear capacitance.





Power losses in power MOSFETs (I).



- **Static losses:**

- Reverse losses \Rightarrow negligible in practice due to the low value of the drain current at zero gate voltage, I_{DSS} .

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I_{DSS}	Zero gate voltage drain current	$V_{DS} = 100 \text{ V}; V_{GS} = 0 \text{ V}$ $V_{DS} = 80 \text{ V}; V_{GS} = 0 \text{ V}; T_j = 175^\circ\text{C}$	-	0.05	10	μA
			-	-	250	μA

- Conduction losses, due to $R_{DS(on)}$:

$$P_{MOS_cond} = R_{DS(ON)} \cdot I_{D_RMS}^2,$$

where I_{D_RMS} is the RMS value of the drain current.

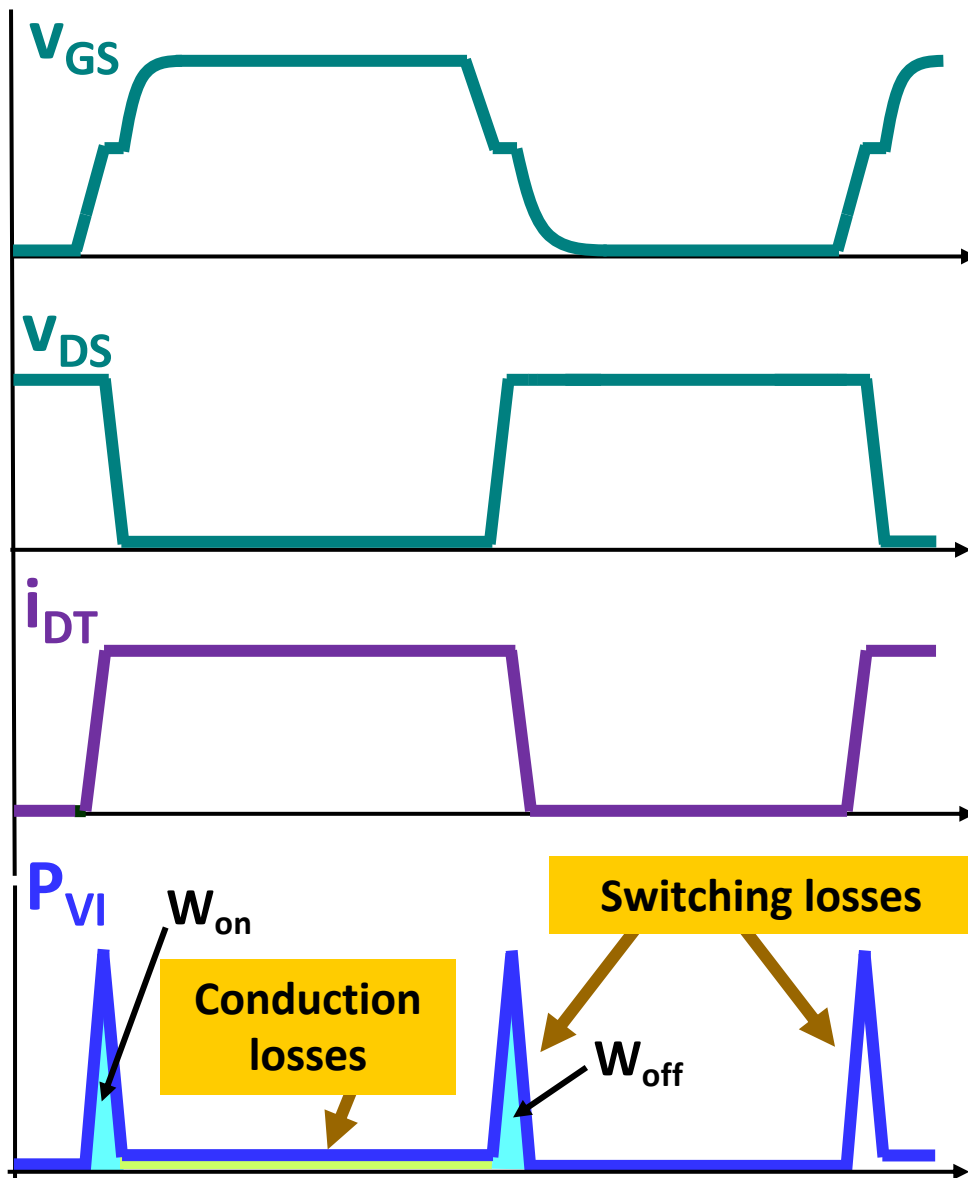
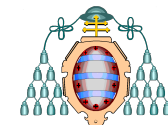
- **Switching (dynamic) losses:**

- Turn-on losses and turn-off losses.

- **Driving losses.**

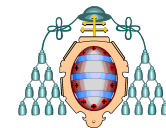


Power losses in power MOSFETs (II).

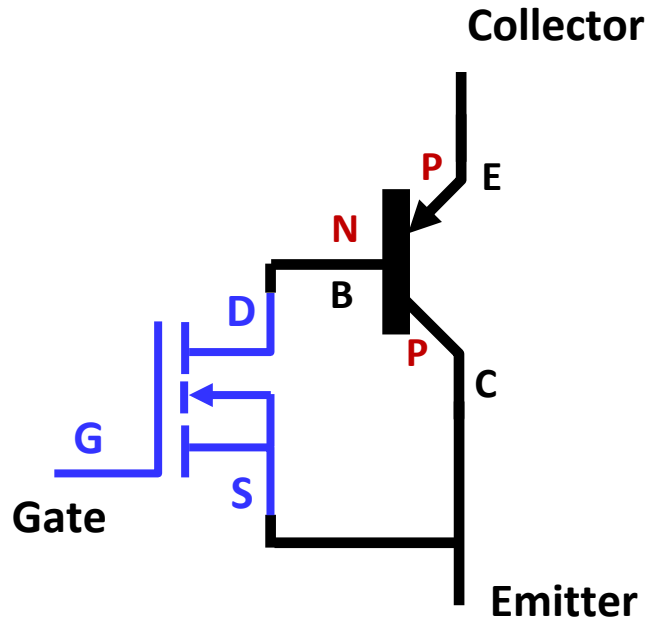


$$P_{MOS_cond} = R_{DS(on)} \cdot I_{D_RMS}^2$$

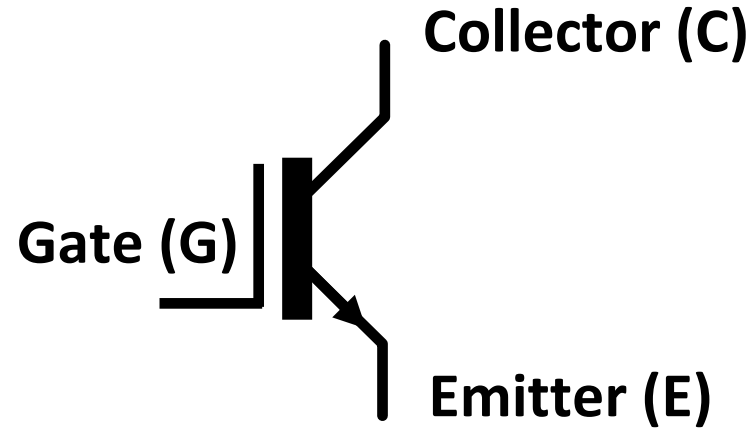
$$P_{MOS_s} = f_s (w_{on} + w_{off})$$



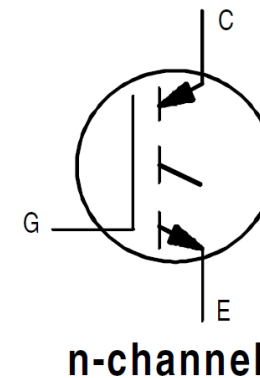
IGBT



Simplified equivalent circuit for an IGBT.



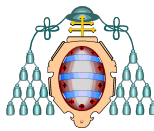
Schematic symbol for a N-channel IGBT.



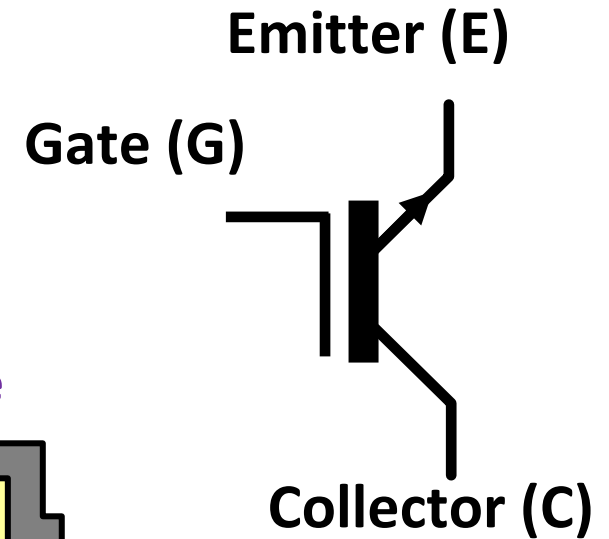
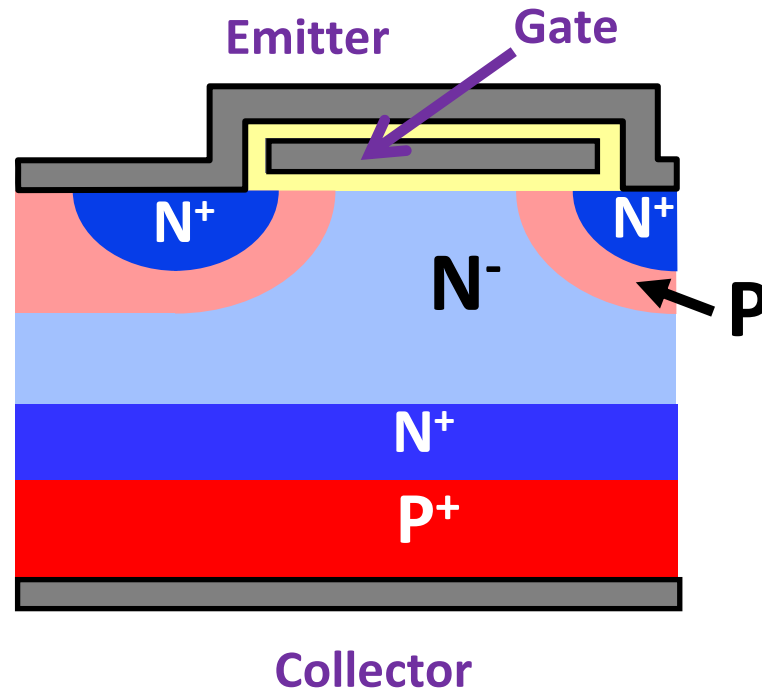
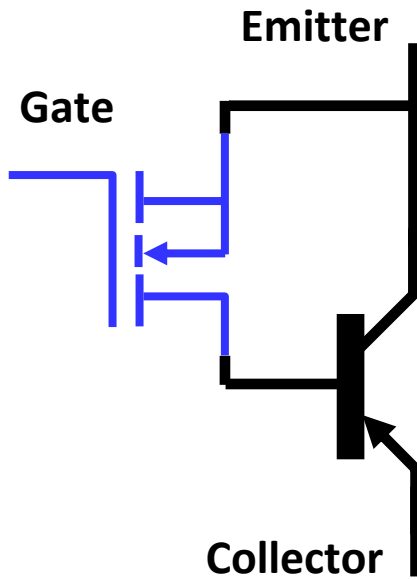
Another schematic symbol also used.



Principle of operation and structure of the IGBT (III).

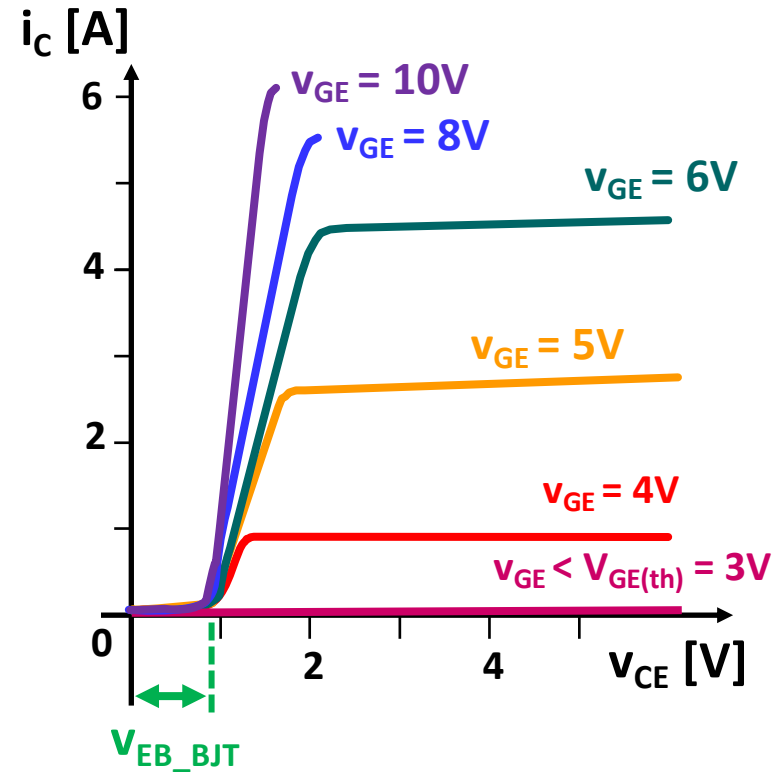
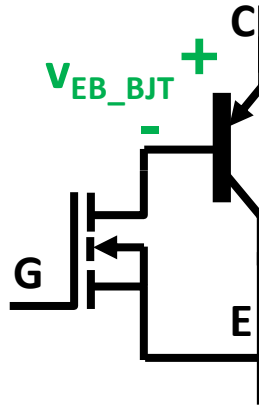
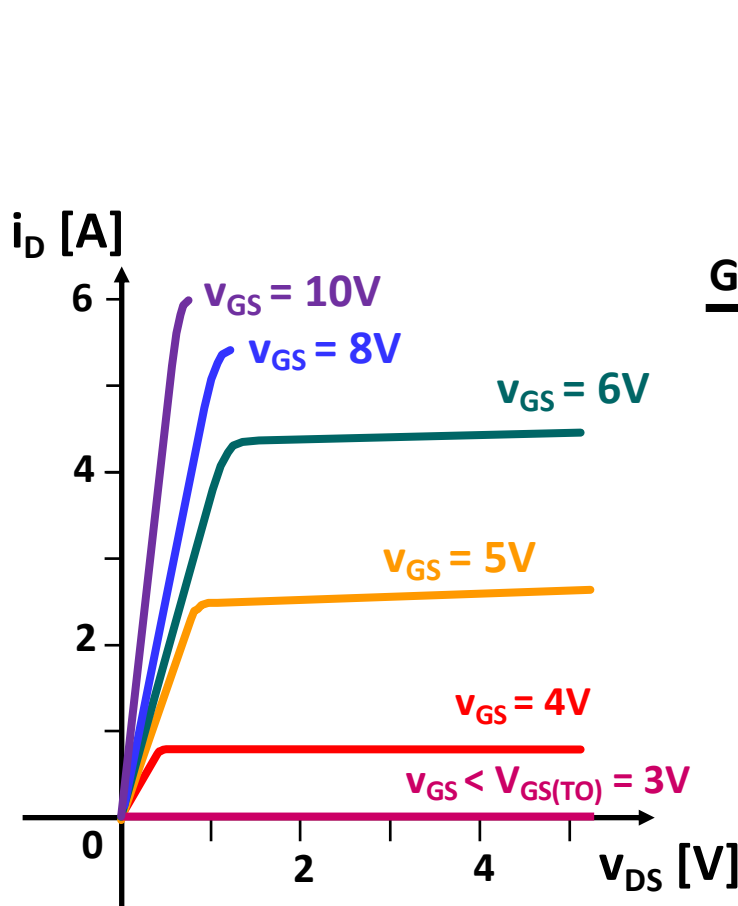
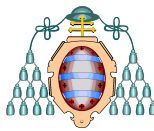


- Internal structure (I).





Static output characteristic curves of a IGBT.

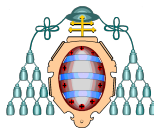


- Static output characteristic curve of a MOSFET.
- It is also the one corresponding to the MOSFET part of a IGBT.

- Static output characteristic curve of a IGBT.
- It can be easily obtained from the MOSFET characteristic curve by adding the voltage drop v_{EB_BJT} corresponding to the emitter-to-base junction of the BJT part of the IGBT.



General characteristics of the IGBTs (I).

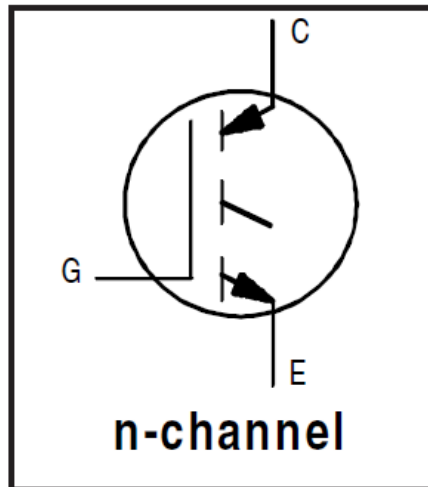


- We will use a specific IGBT to address the general IGBT characteristics.

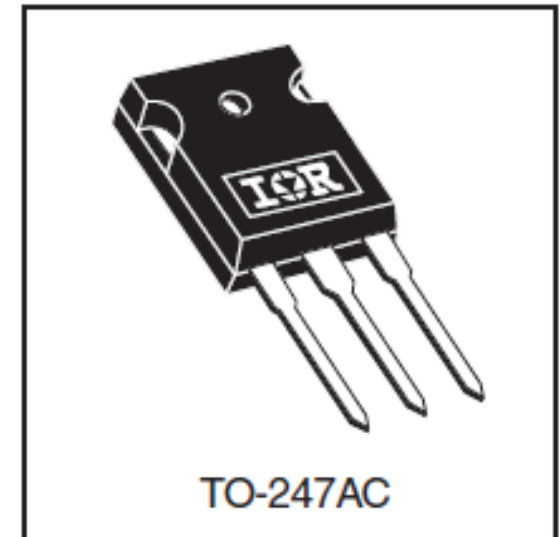
International
IR Rectifier

INSULATED GATE BIPOLAR TRANSISTOR

IRG4PC50W

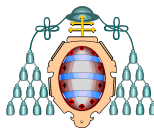


$V_{CES} = 600V$
 $V_{CE(on) \text{ max.}} = 2.30V$
@ $V_{GE} = 15V, I_C = 27A$





General characteristics of the IGBTs (II).



• General information regarding the IRG4PC50W.

Features

- Designed expressly for Switch-Mode Power Supply and PFC (power factor correction) applications
- Industry-benchmark switching losses improve efficiency of all power supply topologies
- 50% reduction of E_{off} parameter
- Low IGBT conduction losses
- Latest-generation IGBT design and construction offers tighter parameters distribution, exceptional reliability

Benefits

- Lower switching losses allow more cost-effective operation than power MOSFETs up to 150 kHz ("hard switched" mode)
- Of particular benefit to single-ended converters and boost PFC topologies 150W and higher
- Low conduction losses and minimal minority-carrier recombination make these an excellent option for resonant mode switching as well (up to >300 kHz)

Tiristor (SCR)

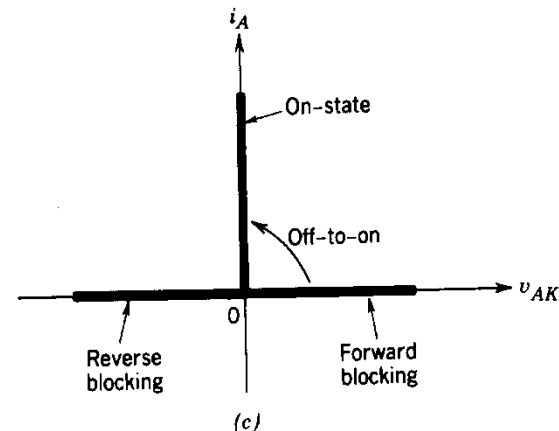
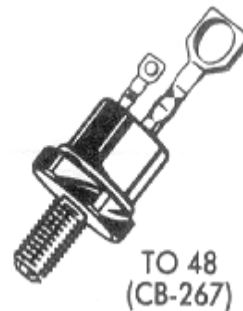
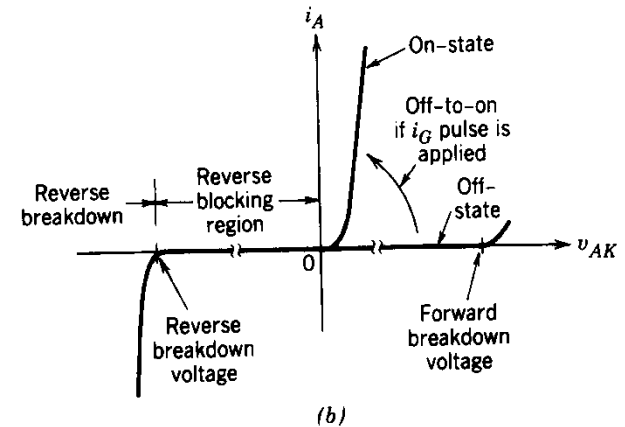
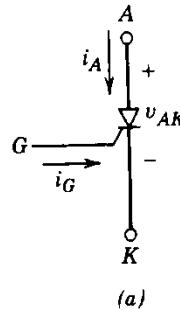
Silicon Controlled Rectifier

Controlable el encendido

Apagado:

- ✓ Natural: la corriente decrece por si sola (rectificadores)
- ✓ Forzado: hacen falta componentes adicionales para bloquear la corriente

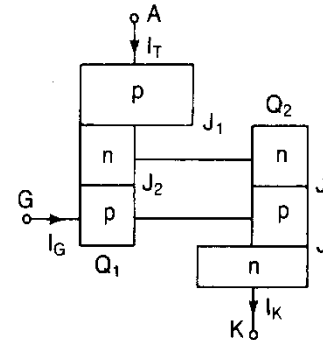
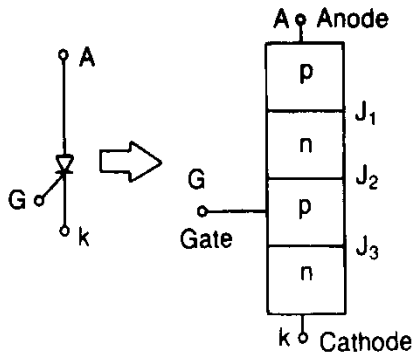
Caída de tensión $\approx 2V$



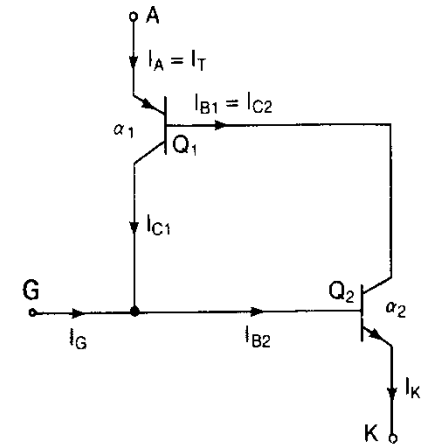


Tiristor (SCR)

Estructura interna y circuito equivalente



(a) Basic structure



(b) Equivalent circuit

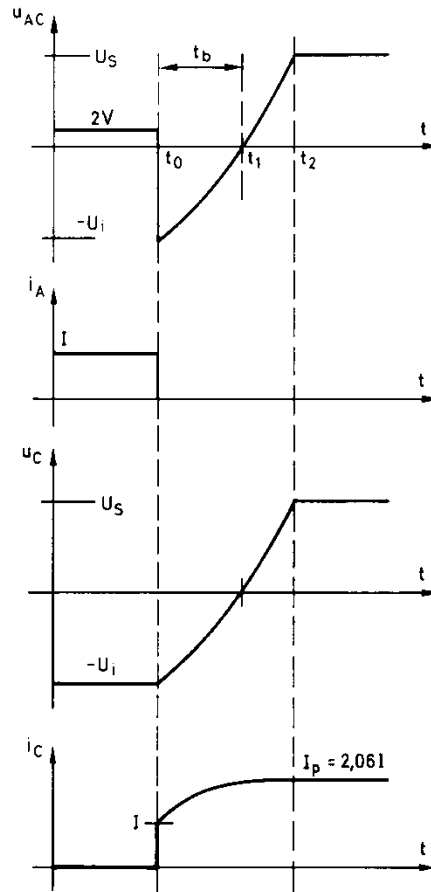
Formas de disparo:

- ✓ Corriente de puerta
- ✓ Luz: light activated thyristors (LASCR)
- ✓ Formas no recomendadas:
 - Tensión alta
 - Estrés térmico
 - Alta derivada de tensión dv/dt

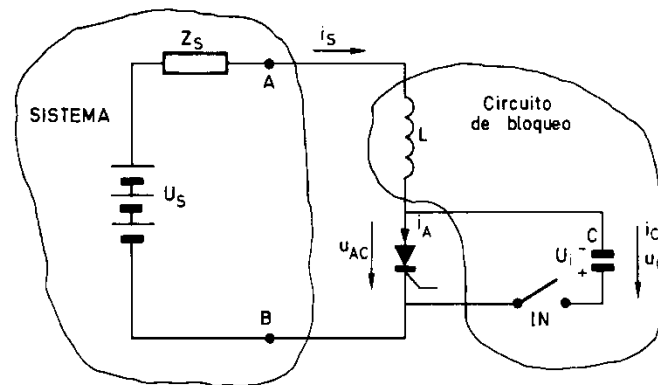


Tiristor (SCR)

Circuito de bloqueo

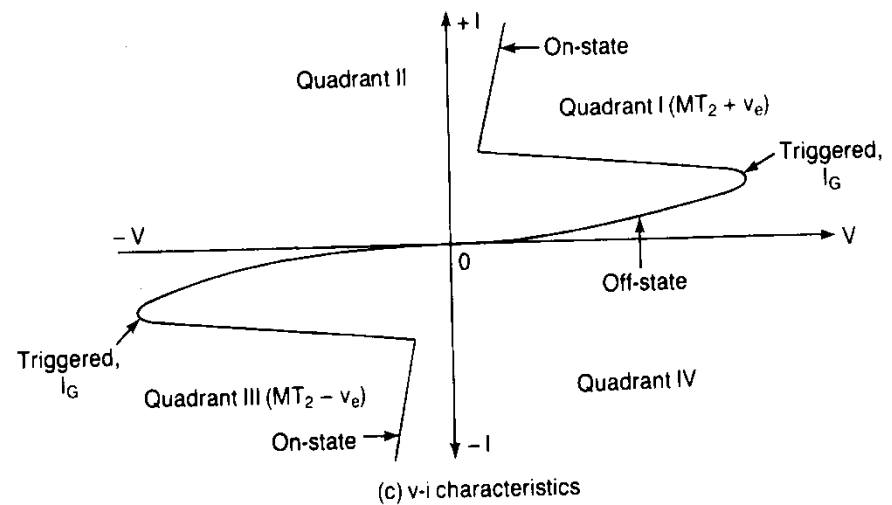
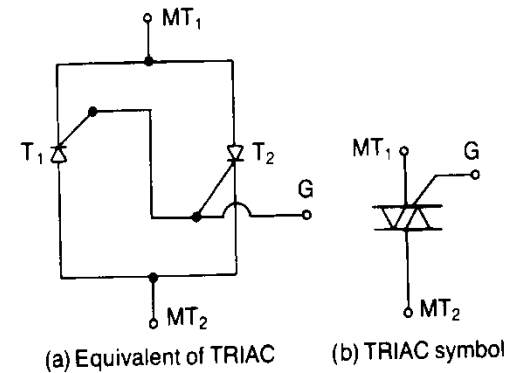
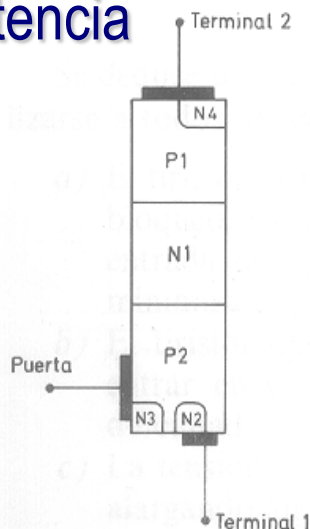


- Tiristor adicional
- Condensador
- Bobina
- Circuito de disparo

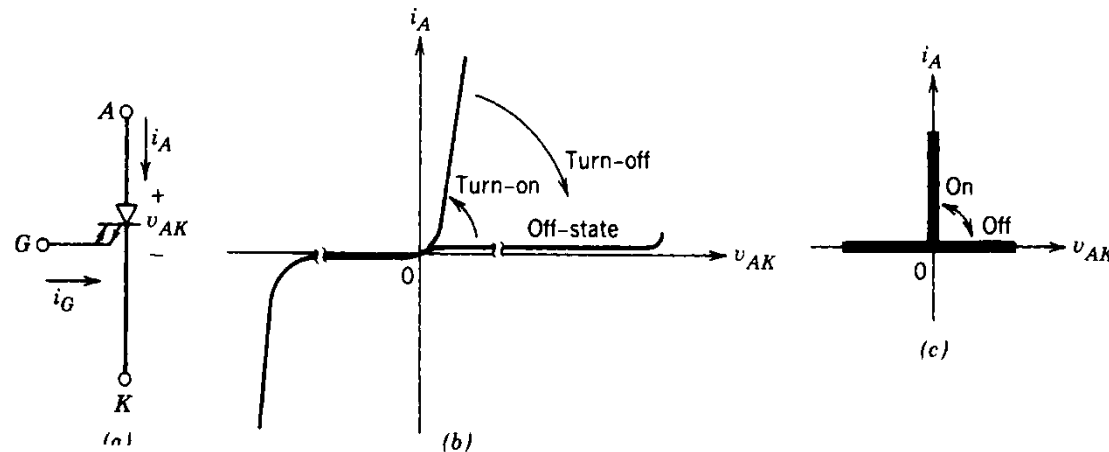


Triac

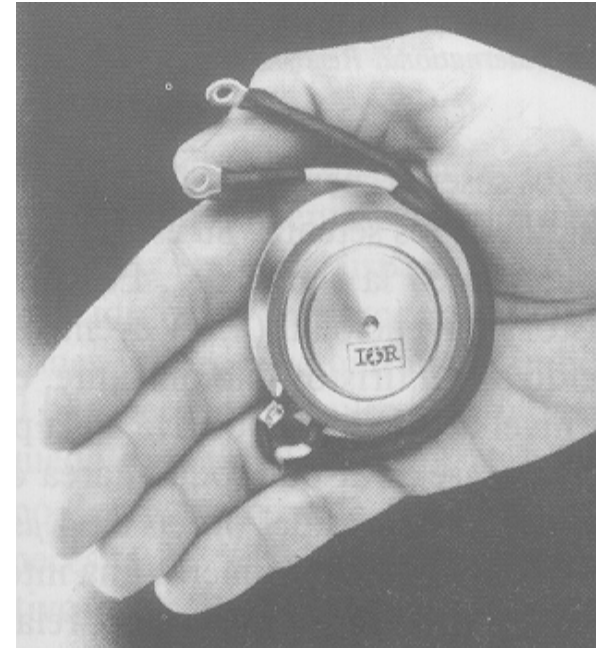
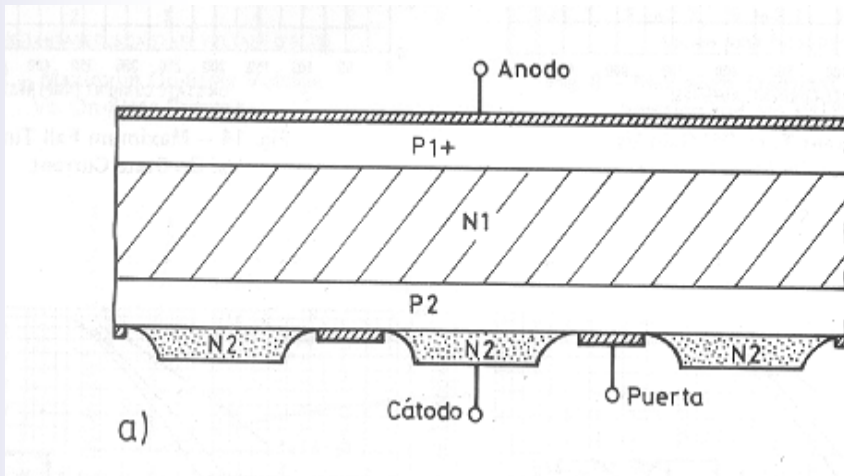
- Controlable el encendido
- Bi-direccional
- Casi simétrico
- Ideal para control de fase en CA
- Aplicaciones de baja frecuencia y potencia



Gate Turn-Off thyristor



- ⚡ Encendido y apagado controlado por puerta
- ⚡ Necesita gran corriente para apagarlo por puerta
- ⚡ Mayor frecuencia que el SCR (pocos kHz)
- ⚡ Aplicación típica: inversores de alta potencia

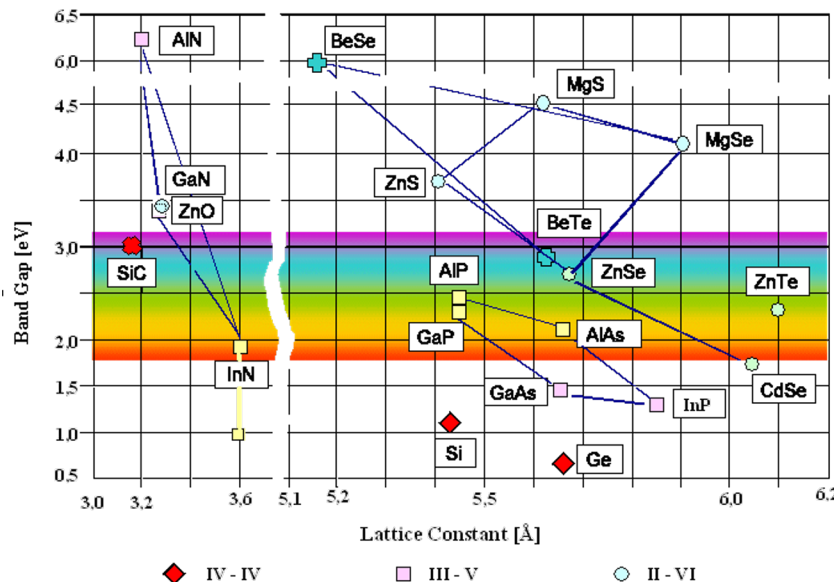
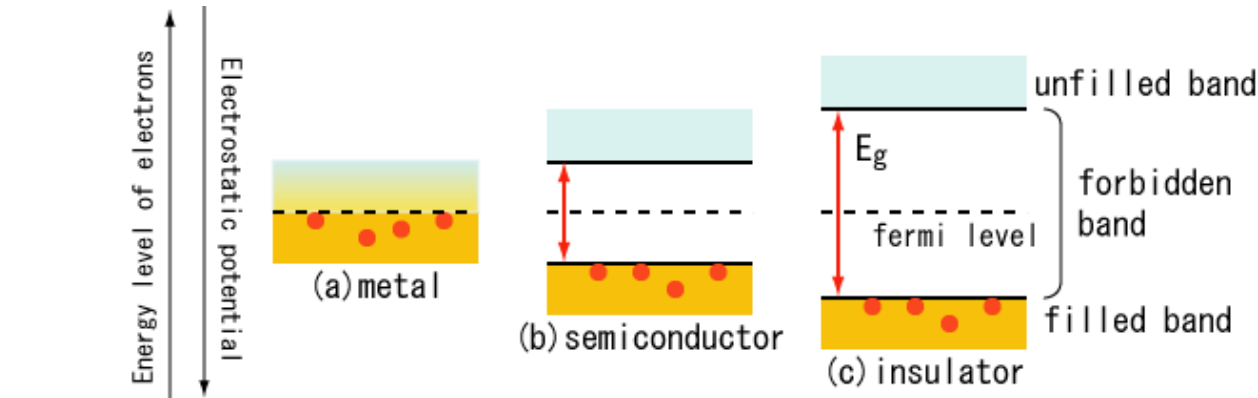


- ✎ Estructura interna similar al SCR pero con mayor influencia de la puerta
- ✎ Mayor caída de tensión en conducción (3V) que el SCR y dependiente de la corriente de puerta

Wide bandgap semiconductors

How do we define them?

- Wide bandgap → electronic band-gaps significantly higher than 1 eV (Si has a gap of 1.1 eV)



- We will see a little bit of physics...
- ...and some applications, of course
- Two devices will be considered:
 - SiC MOSFET
 - GaN HEMT

Impact of new SiC and GaN semiconductors in power conversion

Centro de Electronica Industrial

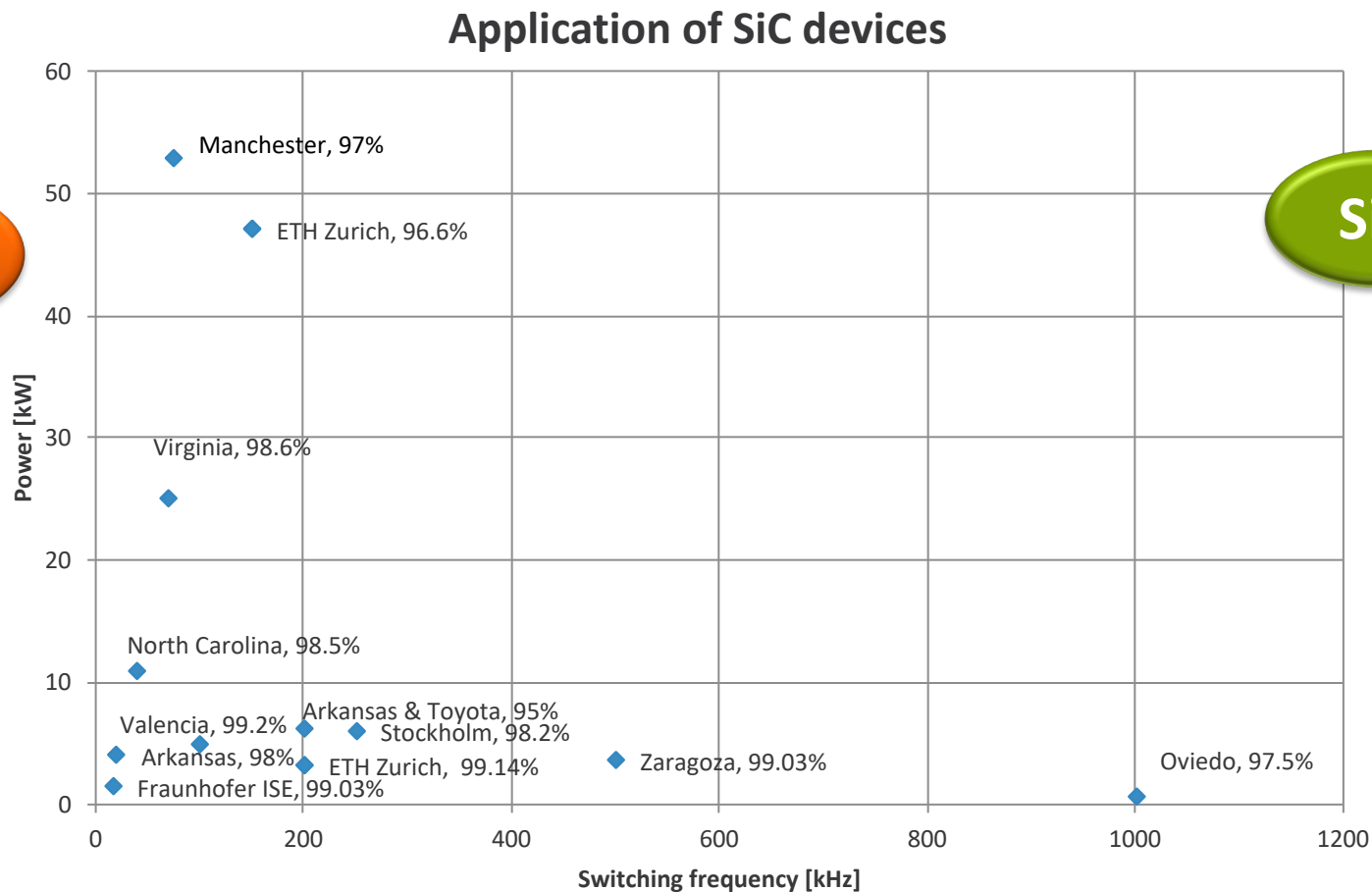
- Para la misma tensión, el SiC es **220 veces menos resistivo**
- Para la misma resistencia, la de SiC soporta **15 veces más tensión**

- **Tecnologías no maduras:**
 - Crecimiento de obleas
 - El sustrato se combe
 - Efectos piezoeléctricos
- El **encapsulado** limita algunas de las ventajas
- Mucho **más caros**



3.

Estado del arte

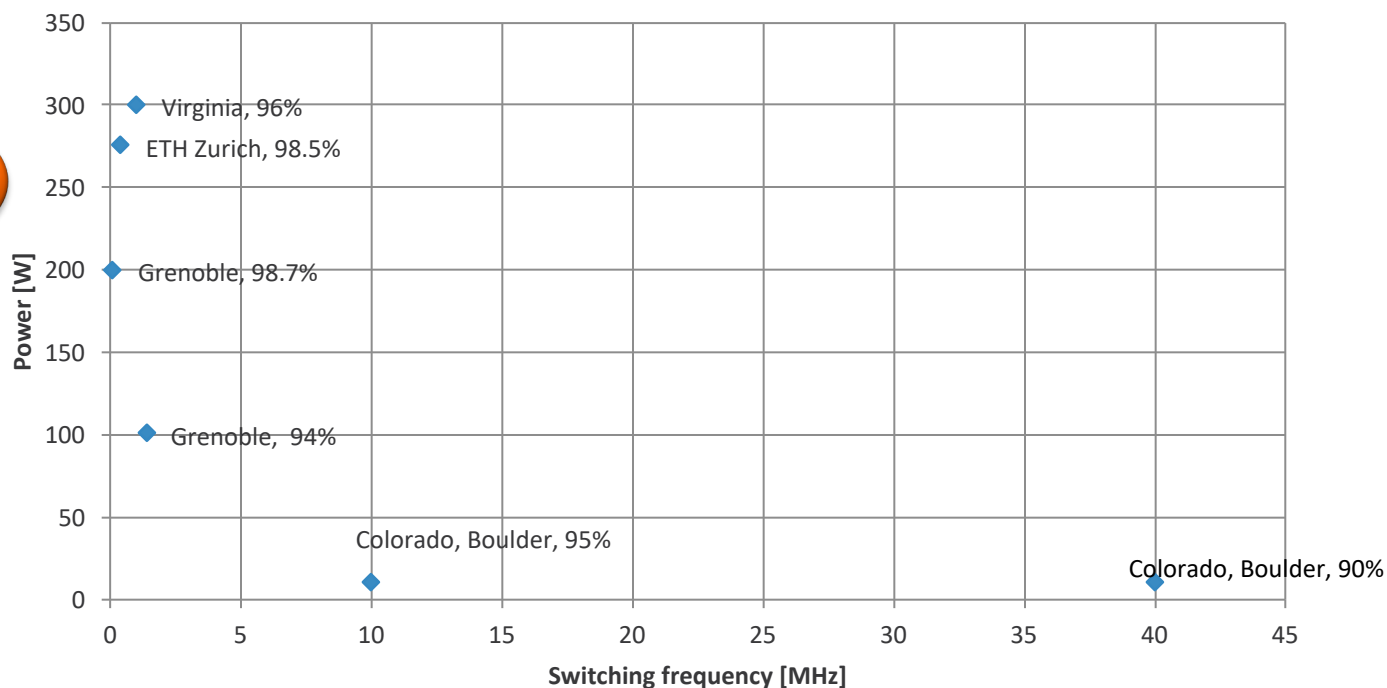


3.

Estado del arte

GaN

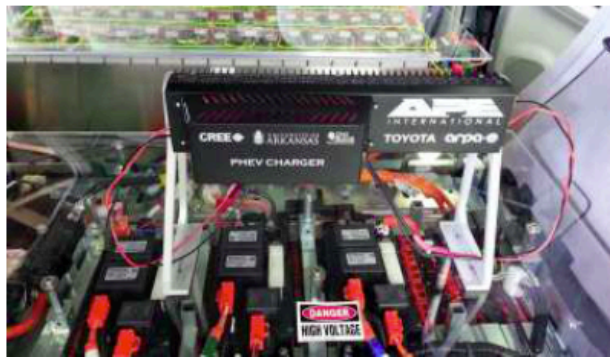
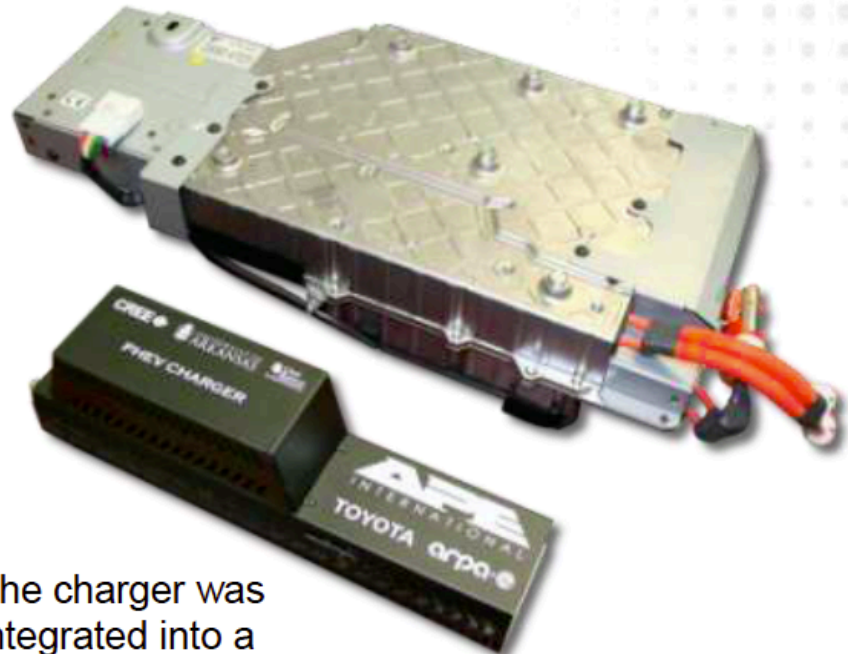
Application of GaN devices



4. Ejemplos del IEEE APEC 2014 (Tim Heidel ARPA-E)

SiC Galvanically-Isolated Vehicle Battery Charger Provides 10x Increase in Power Density

Parameter	Baseline	Prototype
Volumetric Power Density	7.4 W/in ³ (387.5 in ³)	83.3 W/in ³ (73.2 in ³)
Gravimetric Power Density	0.44 kW/kg (6.6 kg)	3.8 kW/kg (1.6 kg)
Power	2.88 kW	6.1 kW
Efficiency	-	95% peak



The charger was integrated into a 2010 model Toyota Prius Plug-in Hybrid

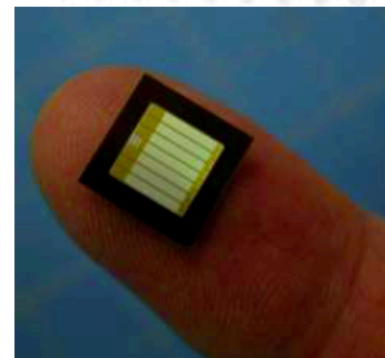
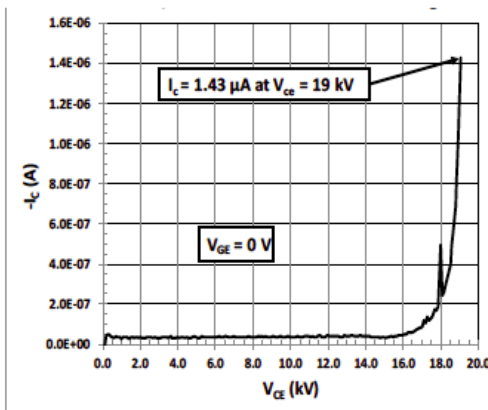
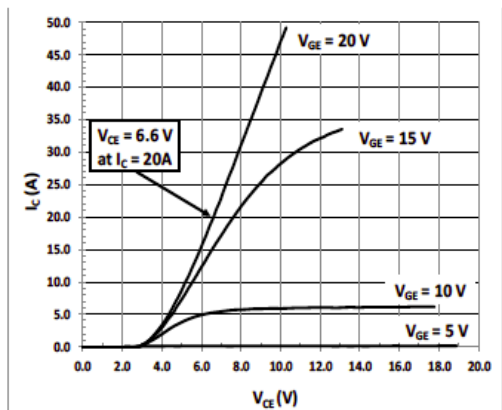
10x Increase in Power Density and Increased Efficiency!

As shown at booth #544.

4.

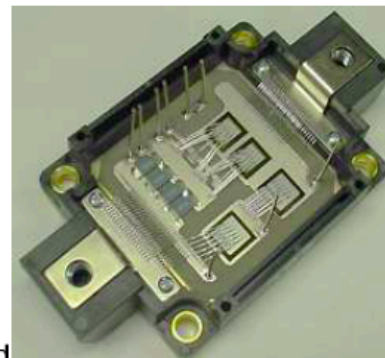
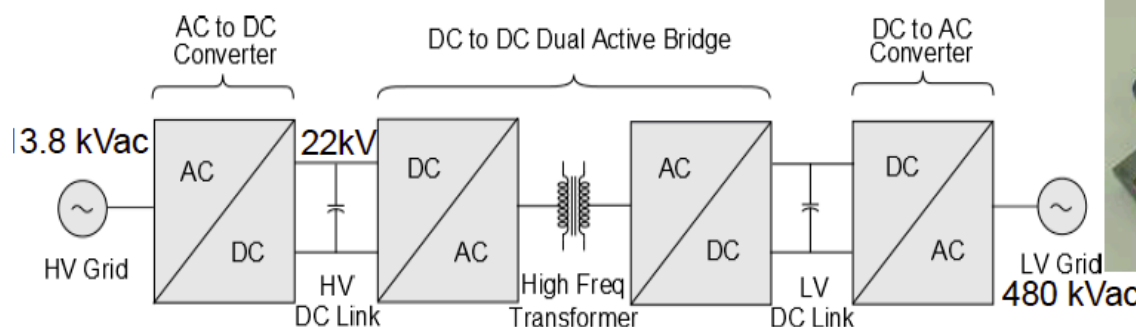
Ejemplos del IEEE APEC 2014 (Tim Heidel ARPA-E)

15kV+ SiC IGBTs



19 kV/20 A SiC n-IGBT

Transformerless Intelligent Power Substation (TIPS)

15kV/40A
SiC IGBT Copack

Conclusiones

- Los nuevos dispositivos de SiC y GaN empiezan a ser una alternativa real sobre los de Si.
- De manera comercial sólo hay algunos dispositivos disponibles:
 - SiC: diodos, JFETs, MOSFETs (típicamente de 1200V)
 - GaN: HEMTs de 100 y 650V
- La tecnología no está madura pero avanza muy rápido.
 - Hay prototipos de tensiones más altas
 - Hay prototipos de IGBTs y dispositivos bidireccionales

2 Relevant applications

Where could GaN play a role?

Potential target for GaN HEMT (non Auto)

